The measuring instruments and auxiliary equipment required are listed in Table 5-17.

### 5.2 Failure Location

In the following test, instructions are given for locating typical possible trouble sources. See also section 2.3 .20 which describes the error messages that can be produced by the software of the Test Receiver ESH 3 when it identifies certain trouble sources.

Section 5.3 contains adjustment procedures and instructions for checking the various boards. For checking the performance specifications of the receiver refer to section 3 .

### 5.2.1 No Response at Switch-on or Readout Makes no Sense

a) Failure symptoms:

At switch-on, the 13-digit display, frequency indication, RF attenuation and demodulator operating range indication, the analog indication and the individual LEDs associated with the device functions do not light up, the readouts obtained make no sense or there are combinations that make no sense, or there is no response to the push of a key.
b) Locating the fault:

- Check the LEDs which monitor the supply voltages of the Analog Power Supply.
- Check the following plug-and-socket connections:

ST9 (Computer board/Analog Power Supply),
BU3 (Computer board/Display Unit) and
ST15 (Motherboard/Analog Power Supply)
BU4/ST3 (Computer board/Memory board).

- Check the voltage on the Computer board:

| ST9 | $3 \mathrm{a}, \mathrm{b} / 4 \mathrm{a}, \mathrm{b} / 5 \mathrm{~b}$ | $5 \mathrm{~V}+5 \%$ |
| :--- | :--- | :--- |
| ST9 | $6 \mathrm{a}, \mathrm{b}$ | +12.0 V |
| ST9 | 7 a | +10.0 V |
| ST9 | 7 b | -10.0 V |

- Check the voltage on the Display Unit:

| ST3 | 9 to 13 | +5 V |
| :--- | :--- | :--- |
| ST3 | 14 | -10 V |
| ST3 | 15 to 16 | +12 V |

- Check the links B1 to B5 on the Computer board and the shorting link STI on the Display Unit.
- Check the microprocessor on the Computer board:

| RESET input | $\mathrm{ST16/3}$ and $\mathrm{B} 2 / 36$ | HIGH potential |
| :--- | :--- | :--- |
| CLOCK output | $\mathrm{B} 2 / 37$ | $3-\mathrm{MHz}$ TIL level |
| RD and WR outputs | $\mathrm{B} 2 / 32$ and $\mathrm{E} 2 / 31$ | Pulses, depending <br> on program |

and carry out signature analysis testing as described in 5.3.1.2.

- Check the microprocessor on the Display Unit:

| RESET input | $S T 2 / 7$ and $B 67 / 4$ | HIGH potential |
| :--- | :--- | :--- |
| PROG. input | $B 67 / 25$ | Pulses depending <br> on program |

and carry out signature analysis testing as described in 5.3.1.1.

### 5.2.2 Receiver Fails to Selfocalibrate

a) Failure symptom:

After the CAL. button 15 has been pressed, the calibration process is interrupted by an error message (time exceeded).
b) Locating the fault:

$$
\begin{array}{ll}
\text { Settings on the ESH 3: } & \text { Indicating mode } 35: A V . \\
& \text { operating range } 33: 20 \mathrm{~dB}
\end{array}
$$

The supply voltages and $500-\mathrm{Hz}$ and $100-\mathrm{kHz}$ reference frequencies must be checked.

- Check the level and the frequency in the TWOPORT mode 38 . Level into $50 \Omega$
$-27 \mathrm{dBm} \pm 1 \mathrm{~dB}$
- Check the level and the frequencies at the four inputs of the calibration generator Y1O as follows:

Inputs:

$$
31.5 \mathrm{kHz}
$$

ST4: $\quad 31.0 \mathrm{kHz}$ (depending on demod. mode) ... $-15 \mathrm{dBm}( \pm 1 \mathrm{~dB})$ 30.0 kHz 28.5 kHz
 66.0015 MHz

ST6: $\quad 66.0000 \mathrm{MHz}$ $-15 \mathrm{dBm}(+1 \mathrm{~dB})$ 65.9985 MHz

ST3: Receiver frequency $+75 \mathrm{MHz} \ldots \ldots \ldots \ldots \ldots$. $15 \mathrm{dBm}( \pm 1 \mathrm{~dB})$
Outputs:
ST2:
Receiver frequency .......................... -27 dBm ( +0.3 dB )
ST1: Receiver frequency .......................... $-67 \mathrm{dBm}( \pm 0.3 \mathrm{~dB})$

- Connect RF socket 45 to GEN. socket 44 with a BNC cable. Set RF attenuation 40 , 41 to 10 dB :

Attenuator (Y16) BU2 ....................................... $37 \mathrm{dBm}(\underline{1} \mathrm{~dB})$
Filter 2 ( Y 8 ) SI8 ................................... $-38 \mathrm{dBm}(+1 \mathrm{~dB})$
Mixers $1 / 2$ (Y9) ST 7 ..................................... $-2 \mathrm{dBm}(+3 \mathrm{~dB})$

- Set RF attenuation and IF attenuation 40, 41 to 50 dB and 40 dB , respectively.

RF level at ST5 of Mixer 3 (Y11) ...........................22 dBm ( $\pm 3 \mathrm{~dB}$ )

- Voltage on pin ST1/a2 of the Indication and AF Demodulation board (Y12) .................................. +1 to +2 V

The analog level indication 14 must give about fulloscale defilection (right-hand scale end).

If not, the fault is in the AGC amplifier of the Analog Circuit (Y3) or in the Mxer 3 (Y11).
5.2.3 Receiver Selfocalibrates but Indication Incorrect in AV. Mode

Settings on the ESH 3: Operating mode 38: TWOPORT.
AII other settings same as under 5.2.2.
Level at BU2 of the Attenuator $Y 16$ $-67 d B+0.3 d B$

The fault is in the feedback amplifier on the Analog Circuit board (Y3) (reference).
5.2.4 Receiver Selfmcalibrates but Indication Incorrect in CISPR Mode

Apply 2 mV EMF at receive frequency.
Level indication 13 60 dBuV

If the indication is incorrect, check the following functions:

- Demodulation and weighting of the indicated signal on the Indication and AF Demod. board (112).
- IF bandwidths of Mixer 3 (Y11) and Mixers 1 and 2 (Y9).
- Calibration pulses produced by the Calibration Generator (Y10).

If the indication is correct, check the IF bandwidths of Mixer 3 (YII) and Mixers 1 and 2 (Y9).

### 5.2.5 Error Messages of ESH3 and Possible Causes

ERROR 01
The frequency entered is too high ( $>30 \mathrm{MHz}$ ).
$\rightarrow$ Operating error

## ERROR 02

The frequency entered is too low ( $<0.009 \mathrm{MHz}$ ).
$\rightarrow$ Operating error

## ERROR 03

The error 03 may occur during the brief calibration if the overall gain of the receiver deviates by more than 0.5 dB from the value determined in the course of the last total calibration. By this error message the user is requested to carry out a total calibration. The message does not implicate a device error.

## ERROR 05

Level calibration at 1 MHz during the brief or total calibration cannot be terminated within the internally fixed time (deviation from nominal value: $>6 \mathrm{~dB}$ ).

## Possible causes:

$\rightarrow$ Error of calibration generator, Check as described in sections 5.3.1.10 and 5.2.2.
$\rightarrow$ Error in signal path. Check as described in 5.2.2.

## ERROR 07

This error may occur during total calibration. The error message informs the user of a correction value exceeding the specified tolerance. If level calibration at 1 MHz is not possible, the ERROR 05 message appears.

During calibration in the Talk-only mode, all correction values can be output to a printer via the IEC bus (printer address: Listen Only). All correction factors up to the last valid value are printed out before the calibration is aborted, thus enabling the user to find the cause of error.

Separated by a blank line, the correction values are printed out in the following sequence:

- Correction values for gain at the different bandwidths and correction values for CISPR 1 (f $>$ 150 kHz ) and CISPR 3 ( $f<150 \mathrm{kHz}$; altogether 4 values):
2.4 kHz ,

500 Hz ( 1 kHz for model 56 ),
200 Hz ,
CISPR 1
CISPR 3

- Linearity correction values for $20-\mathrm{dB}$ operating range
( 3 values; abort of calibration if deviation $>6 \mathrm{~dB}$, flashing of level indication if deviation $>3 \mathrm{~dB}$ ).
- Linearity correction values for $40-\mathrm{dB}$ operating range ( 5 values; abort of calibration if deviation $>6 \mathrm{~dB}$, flashing of level indication if deviation $>3 \mathrm{~dB}$ ).
- Linearity correction values for $50-\mathrm{dB}$ operating range (TValues; abort of calibration if deviation $>6 \mathrm{~dB}$, flashing of level indication if deviation $>3 \mathrm{~dB}$ ).
- Mill correction value for $10-\mathrm{kHz} 1 \mathrm{~F}$ bandwidth
(i value; abort of calibration if deviation $>6 \mathrm{~dB}$, flashing of level indication if deviation $>4 \mathrm{~dB}$ ).
- Erequency-response correction values (for frequency and level)

With a deviation of $>6 \mathrm{~dB}$ from the nominal value ( $=0 \mathrm{~dB}$ ), the calibration is aborted; if the deviation is $>4 \mathrm{~dB}$, the indication flashes.

In case of aborted calibration, proceed as follows to find the cause:

If am appropriate printer is not available, the source of error can be found by observing the calithration process.

If ERROR 07 occurs at the beginning of the calibration (receiver at 1 MHz ), carry out the following tests:

- Check the calibration generator:

Settings on the ESH 3: Frequency 1: MHz
Generator: ON
Connect a power meter to the generator output.
$\rightarrow$ Level at generator output: $-27 \mathrm{dBm} \pm 0.3 \mathrm{~dB}$
If an error has been found, check the calibration generator as described in section 5.3.1.10.2.

- Check the gain at all IF bandwidths:

Settings on the ESH 3: Frequency: 1 MHz
Generator: ON
Connect the generator output to the RF input using a short BNC cable.
$\rightarrow$ Select the IF bandwidths successively. The level indication is $0 \pm 6 \mathrm{~dB}$ at every IF bandwidth (abort criterion; ideal value $< \pm 2 \mathrm{~dB}$ ).

In case of an error, check the 1 st +2 nd mixer PCB (5.3.1.9.5 to 5.3.1.9.8) and the 3rd mixer PCB (5.3.1.11.8).

- Check the CISPR indication:

Settings on the ESH 3: Frequency: 1 MHz Generator: OFF Indication mode: CISPR RF attenuation: 40 dB IF attenuation: 30 dB

Apply a sinewave signal ( $1 \mathrm{MHz}, 60 \mathrm{~dB} \mu \mathrm{~V}$ ) to the RF input.
$\rightarrow$ The receiver indicates $60 \mathrm{~dB} \mathrm{\mu V} \pm 6 \mathrm{~dB}$ (abort criterion; ideal value $< \pm 2 \mathrm{~dB}$ ).
Apply the CISPR-1 standard pulse ( $100-\mathrm{Hz}$ pulse frequency) to the RF input (see 3.2.1.4).
$\rightarrow$ The receiver indicates $60 \mathrm{~dB} \mathrm{\mu V} \pm 6 \mathrm{~dB}$ (abort criterion; ideal value $< \pm 2 \mathrm{~dB}$ ).

Set the frequency to 100 kHz and the $R F$ attenuation to 20 dB .
Apply a sinewave signal ( $100 \mathrm{kHz}, 40 \mathrm{~dB} \mathrm{\mu V}$ ) to the RF input.
$\rightarrow$ The receiver indicates $40 \mathrm{~dB} \mathrm{\mu V} \pm 6 \mathrm{~dB}$ (abort criterion; ideal value $< \pm 2 \mathrm{~dB}$ ).

Apply the CISPR-3 standard pulse (25-Hz puise frequency) to the RF input (see 3.2.1.3).
$\rightarrow$ The receiver indicates $40 \mathrm{~dB} \mu \mathrm{~V} \pm 6 \mathrm{~dB}$ (abort criterion; ideal value $< \pm 2 \mathrm{~dB}$ ).

- Check the indication linearity:

Settings on the ESH 3: Frequency: 1 MHz IF attenuation: 40 dB
RF attenuation: 50 dB
IF bandwidth: 200 Hz
Indication range: 20 dB
Increase RF attenuation in steps up to 70 dB .
$\rightarrow$ The indication deviates from the starting value by 6 dB max.
(abort criterion; ideal value $<1 \mathrm{~dB}$ ).

Select $30-\mathrm{dB}$ RF attenuation and $40-\mathrm{dB}$ indication range.
Increase RF attenuation in steps up to 70 dB .
$\rightarrow$ The indication deviates from the starting value by 6 dB max.
(abort criterion; ideal value $<3 \mathrm{~dB}$ ).

Select $10-\mathrm{dB}$ RF attenuation and $60-\mathrm{dB}$ indication range.
Inrease RF attenuation in steps up to 70 dB .
$\rightarrow$ The indication deviates from the starting value by 6 dB max.
(abort criterion; ideal value $<3 \mathrm{~dB}$ ).
In case of an error, check the indication section on the indication and AF demodulation PCB as described in 5.3.1.12.2.

When the calibration is aborted while the frequency response is measured, check the preselection filter on the filter 1 or filter 2 PCB, depending on which filter is in circuit when the abort occurs. The cut-off frequencies of the filters are specified in Tables 5-2 (section 5.3.1.7) and 5-3 (section 5.3.1.8). Check according to 5.3.1.7 and 5.3.1.8.

If no error is found in the input filters, check the oscillator level at input 5 T3 (X3) of the $95 t+2$ nd mixer PCB (nominal value: $+23 \mathrm{dBm} \pm 2 \mathrm{~dB}$ ).

## ERROR 08:

The memory register called up with RCL is not occupied.
$\rightarrow$ Operating error

## Vottage supply errors:

## ERROR 10:

Error in $+10-\mathrm{V}$ supply. The voltage is either not available or not within tolerance. The $+10-\mathrm{V}$ LED on the power supply lights.
$\rightarrow$ Remove power connector from motherboard. If the $+10-V$ LED goes out, one of the PCBs is faulty, if it remains lit, trace the error in the power supply.
$\rightarrow$ In case of a PCB error, remove the PCBs one after the other until the error has disappeared.

## ERROR 11:

Error in $-10-\mathrm{V}$ supply. The voltage is either not available or not within tolerance. The -10-V LED on the power supply lights.
$\rightarrow$ Trouble-shooting analogous to that of ERROR 10.

## ERROR 12:

Errar in $+12-\mathrm{V}$ supply. The voltage is either not available or not within tolerance. The $+12-\mathrm{V}$ LED on the power supply lights.
$\rightarrow$ Trouble-shooting analogous to that of ERROR 10.

## ERROR 13:

Error in $+25-\mathrm{V}$ supply. The voltage is either not available or not within tolerance.
$\rightarrow$ Power consumption too high or short circuit on one of the PCBs synthesizer 1. synthesizer 2 , mixers $1+2$ or analog circuit.
$\rightarrow$ Error in power supply.
Remove the PCBs one after the other until the error has disappeared.

## ERROR 14:

Error in $+30-\mathrm{V}$ supply. The voltage is either not available or not within tolerance.
$\rightarrow$ Power consumption too high or short circuit of filter control PCB.
$\rightarrow$ Error in power supply.

## ERRÓR 20 to ERROR 41:

These errors are operating errors, They are defined in Table 2-6 of the ESH 3 Operating Manual.

## Synthesizer errors:

## ERROR 51:

The 1st oscillator ( 75 to 105 MHz ) on the synthesizer 1 PCB does not lock.
Trouble-shooting:

- Check $65-$ to- $65.1-\mathrm{MHz}_{8}-16 \pm 2-\mathrm{dBm}$ signal at $5 T 2$.
- Check $100-\mathrm{kHz}$ reference at ST1.b18 (TTL level).
- Check $65-\mathrm{MHz}$ amplifier on synthesizer 1 PCB according to section 5.3.1.5.4.
- Check $10-$ to- $40-\mathrm{MHz}$ amplifier on synthesizer 1 PCB according to section 5.3.1.5.5.
- Check the oscillators according to section 5.3.1.5.


## ERROR 52:

The 50 -to- $51-\mathrm{MHz}$ interpolation oscillator on the synthesizer 2 PCB does not lock.

## Trouble-shooting:

- Check setting of int./ext. reference switch on rear panel of receiver.
- Check reference oscillator of synthesizer 2 according to section 5.3.1.4.2.
- Check 50-to-51-MHz oscillator according to section 5.3.1.4.3.


## ERROR 53:

The 2nd oscillator ( 66 MHz ) on the synthesizer 2 PCB does not lock.

## Trouble-shooting:

- Check int./ext. reference switch on rear panel.
- Check reference oscillator of synthesizer 2 according to section 5.3.1.4.2.
- Check $66-\mathrm{MHz}$ oscillator of synthesizer 2 according to section 5.3.1.4.4.


## ERROR 54:

The 3rd oscillator ( 8.97 MHz ) on the mixer 3 PCB does not lock.

## Trouble-shooting:

- Check $500-\mathrm{Hz}$ reference at ST1.bl of 3rd mixer. In case of epror, check synthesizer 2.
- Check 8.97-MHz oscillator of 3rd mixer according to section 5.3.1.11.2.


## ERROR 55:

The $30-\mathrm{kHz}$ oscillator on indication and AF demodulation PCB does not lock.
Trouble-shooting:

- Check $500-\mathrm{Hz}$ reference at STI.abi of indication and AF demodulation PCB. In case of error, check synthesizer 2.
- Check $30-\mathrm{kHz}$ oscillator on indication and AF demodulation PCB according to section 5.3.1.12.2 h.


## Several errors at a time:

If several error messages appear at the same time, they generally have a common source. The following table lists the secondary errors occurring as a result of synthesizer errors.

| Source | Secondary arror |
| :--- | :--- |
| Reference oscillator not available | ERROR 51, 52, 53, 5A,55 |
| ERROR 51 | . |
| ERROR 52 | ERROR 51, 53' |
| ERROR 53 | - |
| ERROR 54 | - |
| ERROR 55 | - |

Table 2-6 Error code list

| 01 | Frequency entered above limit |
| :---: | :---: |
| 02 | Frequency entered below limit |
| 03 | CAL:CHECK Comparison frequency response correction/ current value $\geq 0.5 \mathrm{~dB}$, occurs after an aborted total calibration, for example. |
| 04 | No listener on IEC bus (fault in IEC-bus controller) |
| 05 | Level or offset calibration is not accomplished within fixed time (hardware error) |
| 07 | Correction value at CAL. TOTAL $>6 \mathrm{~dB}$; total calibration aborsed. |
| 08 | Memory register not occupied on RCL |
| $\begin{aligned} & 10 \\ & 11 \\ & 12 \\ & 13 \\ & 13 \end{aligned}$ | $\begin{aligned} & +10 \mathrm{~V} \\ & +10 \mathrm{~V} \\ & +12 \mathrm{~V} \\ & +25 \mathrm{~V} \\ & +30 \mathrm{~V} \end{aligned}$ <br> Failure of a supply voltage (the failure of the $+5-\mathrm{V}$ supply voltage is not indicated) |
| $\begin{aligned} & 20 \\ & 21 \\ & 22 \\ & 23 \\ & 24 \\ & 25 \end{aligned}$ | Current register <br> Register 1 <br> Register 2 <br> Register 3 <br> Register 4 <br> Register 5 <br> At start of automatic frequency scan, one or more values are not defined. |
| $\begin{aligned} & 30 \\ & 31 \end{aligned}$ | START frequency > STOP frequency <br> START frequency $=\begin{gathered}\text { STOP frequency and } X Y \text { recorder or } Z S G 3 \\ \text { connected }\end{gathered}$ |
| 32 | MAX. PEGEL S MIN. PEGEL |
| 33 | SPEC.FUNC. 61 X axis logarithmic and fstop: fitart < 1.4 |
| 40 | ZSG 3 error: <br> Error message if SPEC.FUNC. 61 is selected for SCAN RUN with ZSG 3 |
| 51 | Synthesizer 1: 1st oscillator has not locked |
| 52 | Synthesizer 2: 50-to-51-MHz oscillator has not locked |
| 53 | Synthesizer 2: 2nd oscillator has not locked |
| 54 | Mixer 3: 3rd oscillator has not locked |
| 55 | Indication and AF demodulation: $30-\mathrm{kHz}$ oscillator has not locked |

### 5.3.1 Adjustment of the Various Boards

To facilitate measurements on and adjustment of the boards normally plugged directly into the motherboard of the ESH 3. Rohde \& Schwarz supply the Service Unit ESH2-Z7 (IN 338.4112) which permits operation of these boards outside of the receiver.
5.3.1.1 Display Unit
5.3.1.1.1 Supply Voltages

| Pin | Voltage | Current |
| :--- | :--- | :--- |
| ST3.1 to .7 | Ground | - |
| ST3.9 to .13 | +5 V | max. 2 A |
| ST3.15 to .16 | +12 V | approx. 50 mA |
| ST3.14 | -10 V | approx. -10 mA |

### 5.3.1.1.2 Frequency Indication

- Enter BCD-coded frequency information at ST1.1 to ST1.23.
- Check the resulting frequency indication (B18-B23).


### 5.3.1.1.3 Analog Level Indication

- Apply analog voltage GND to ST3.8 and 0.355 V to 3.55 V to ST2.12.
- Adjust R51
for the first (left-hand) IED (GI40) to light up at 0.355 V and the last (right-hand) LED (GL43) to light up at 3.55 V .
- Adjust R47
until the intensity of the right-hand LED array (GI42, GL43) is the same as that of the left-hand IED array (GI40, GL41).
- Apply a logic I to ST1. 4 and ST2.8 and check the MIN. LED (GI38) and the MAX. IED (GL39) respectively.


### 5.3.1.1.4 Analog Offset Indication

- Apply analog voltage GND to ST3.8 and 1.0 V to 4.0 V to $S T 1.24$.
- Adjust RI
for the two centre LEDs (GI44, GI45) to light up at 2.5 V .
- Apply a logic L to ST1. 16 and check the FREQUENCY OFFSET LED (GI88).


### 5.3.1.1.5 Signature Analysis

The microcomputer B67 contains a signature analysis program which can be activated by changing the connection of the link ST4, and which permits checking the digital driving circuit for the $5 \times 7$ dot matrices and all
the LEDs which indicate the device settings. First check visually that all the LEDs and dot matrices light up.

To do so:

- Change the connection of Link ST4.
- Reset the processor B67 by momentarily connecting ST7. 1 to ground.
- Check the clock pulses at SI7.3:

- Connect the signature analyzer to ST5 and set as follows:


Subsequently check the board against the above diagram and signature analysis table.

### 5.3.1.1.6 Keyboard and Display Interface

The keyboard and display chip $B 61$, which is a selfocontained unit, is driven directly from the main processor on the Computer board and is covered by the test and signature analysis of the latter. The test is therefore carried out in conjunction with the Computer board and the relevant signature analysis table (section 5.3.1.2.2.1).

### 5.3.1.1.7 Difference Current Sink

- Switch on signature analysis in accordance with 5.3.1.1.5 (all the LEDs and dot matrices Iight up).
- Connect oscilloscope to ST3.1 to 7 (GND) and ST3.9 to 13 (+5V) and watch screen display.


### 5.3.1.2.1 Supply Voltages

| Pin | Voltage | Current |  |
| :--- | :--- | :--- | :--- |
| ST9 | $1 \mathrm{a} / 1 \mathrm{~b} / 2 \mathrm{a} / 2 \mathrm{~b}$ | Ground | - |
| ST9 | $3 \mathrm{a} / 3 \mathrm{~b} / 4 \mathrm{a} / 4 \mathrm{~b} / 5 \mathrm{a}$ | +5 V | approx. 2 A |
| ST9 | $6 \mathrm{a} / 6 \mathrm{~b}$ | +12 V | approx. 50 mA |
| ST9 | 7 a | +10 V | approx. 100 mA |
| ST9 | 7 b | -10 V | approx. 100 mA |

### 5.3.1.2.2 Signature Analysis

Free-running and software-driven signature analysis is used in fault location when testing the Computer board, which constitutes a complex digital system. With free-running signature analysis, the processor must cyclically scan its entire address range, and the start/stop signal for the signature analyzer originates directly from the address bus.

With software-driven signature analysis, a cyclic control program for all the peripheral chips is stored in the EPROMs, making for more thorough individual testing.

### 5.3.1.2.2.1 Free-running Signature Analysis

An initial but effective test of the Computer board can be carried out using free-running signature analysis. In this signature analysis mode, the feedback between the processor and the EPROMs memory is interrupted (bus driver B6), and driver $B 83$ is activated, which produces a NOP instruction in every read cycle of the processor. The signatures are used for checking the processor and the EPROMs as well as the driver chips and the address decoder chips.

Settings and checks:

- To select the free-running mode, change over the connection of the links $B R 2$ and $B R 5$ to the position shown: $B R 2 \square$ BR5 $\square \square$.
- Reset the processor by: momentarily applying a logic $f$ to ST16.2 (normally at $H$ level); or switching the unit off.
- Check the CLOCK output B2. 37 of the processor: $3-\mathrm{MHz}$ clock signal at TML level.
- Check the signatures of the address bus.
- Connect the signature analyzer to STI 4.

Settings on the 5004A:
$\operatorname{START} \square, \operatorname{STOP} \square$
Signatures B2.21 to B2.28 ( A A8 to A15) and B5.4. B5.6 to B5.21 ( $\approx$ AO to A7):

| AO | UUUU | A8 | HC89 |
| :--- | :--- | :--- | :--- |
| A1 | 5555 | A9 | $2 H 70$ |
| A2 | CCCC | A10 | HPPO |
| A3 | $7 F 7 F$ | A11 | 1293 |
| A4 | $5 H 21$ | A12 | HAP7 |
| A5 | OAFA | A13 | $3 C 96$ |
| A6 | UPFH | A14 | 3827 |
| A7 | $52 F 8$ | A15 | $755 U$ |

Check these signatures on the chips B1, B4, B7 as well as on the EPROMs B8, B11, B14, B15 and the socket BU4 on the Memory board.

- Check the signatures of the data bus.

Connect signature analyzer to ST13 (contents of EPROMs).
Settings on the 5004 A :
$\operatorname{START} \square$ STOP $L$ CLOCK

All the signatures are listed in the table given in the Appendix. After the test has been accomplished, reconnect the links BR2 and BR5 in the original positions.

### 5.3.1.2.2.2 Software-driven Slgnature Analysis

The next step in testing the Computer board is a special test program provided in the EPROMs:

- Change the connection of Iink BRI.
- Reset the processor by momentarily applying a logic $L$ to ST16.3.
- Check the signatures of the peripheral chips against the table given in section 5.3.1.2.2.1.
- Check the staircase signals ( 0 V to +10 V ) at the two $D / A$ converter outputs via $\mathrm{B}_{4} 1$, on the oscilloscope.

After the test has been accomplished, reconnect the link BRI in the original position.

### 5.3.1.2.3 A/D Converter

All the voltages to be measured or applied are with respect to $S T 9.10 \mathrm{~b}$.
A special test routine of the processor software can be used for adjustment of the A/D converter B26, B27. For this purpose, EPROMs B8, B11, B14, B15 must be fitted, and the Memory board must be connected to BU4, just as for the signature analysis.

By changing the connection of the link BR6 and resetting the processor (logic L at ST16.3), the A/D converter continually receives start-of-conversion pulses, and its output is continuously read out.

- Adjust the reference voltage at MP3 to $+5.000 \mathrm{~V}(+1 \mathrm{mV})$ by means of R6.
- Adjust the sample/hold amplifier B77:

Connect the sample/hold input ST5.22 to ST9.10b. Adjust the voltage at MP6 to $0 \mathrm{~V}(\underline{1} \mathrm{mV})$ by means of R12.

- Adjust the $A / D$ converter B26:

Apply a voltage of $+5.000 \mathrm{~V}(\underline{1} \mathrm{mV})$ to input ST5.22. Adjust the gain of the $A / D$ converter by means of R5 such that the logic level at B30.1 is just at the switching point between $H$ and $L(=A / D$ converter value 1023). The value of the $A / D$ converter is output directly to the 13 -digit display of the Display Unit. After the adjustment has been accomplished, reconnect link BR6 in the original position.
5.3.1.3 Analog Circuit Board
5.3.1.3.1 Supply Voltages

|  | Pin | Voltage | Current |
| :--- | :--- | :--- | :--- |
| ST1 | 22 a | +10 V | approx. 50 mA |
|  | 22 b | -10 V | approx. -10 mA |
|  | 23 b | +5 V | approx. 25 mA |
|  | 24 b | GND |  |
| 23 a | +12 V | $0 / 7.5 \mathrm{~mA}$ with AF off $/ \mathrm{AF}$ on |  |
|  | 24 a | +25 V | $<1 \mathrm{~mA}$ |

### 5.3.1.3.2 General

All analog switching functions are performed by means of CMOS switches, a logic $H=+10 \mathrm{~V}$ at the control input of the CMOS switch corresponding to $O N$ and a logic $L=<1 \mathrm{~V}$ corresponding to OFF.

For driving the required inputs (Table 4 m ), TTL levels must be applied which are converted to 0 or 10 V by means of level shifters.

If switches off, the resistance of the CMOS switches is $>10 \mathrm{M} \Omega$, and if switched on approx. $300 \Omega$.

NOTE: Analog DC voltage values given assume a supply voltage of $+10 \mathrm{~V}+1 \%$.

### 5.3.1.3.3 Analog Level Path

### 2.3.1.3.3.1 Level Calibration

- After the supply voltage has been applied, apply a variable voltage ( 0.2 to $2 \mathrm{~V})$ to $\mathrm{STl} / 17 \mathrm{a}$.
- Measure the voltage at B1/III pin 3 by means of a digital voltmeter. $\mathrm{V}_{\text {center }}=2 \mathrm{~V}$ with a tolerance of $\pm 0.02 \mathrm{~V}$. While the input voltage being measured is at $V_{\text {center }}+20 \mathrm{mV}$, a +5 V signal ( H level) must appear at ST1/21b signifying calibration complete.
- If $\mathrm{V}_{\text {in }}<\mathrm{V}_{\mathrm{m}}$, the output of the comparator B1/III pin 1 should be approx. +9 V and if $\mathrm{V}_{\text {in }}>\mathrm{V}_{\mathrm{m}}$ it should be approx. -5 V .
- If the logic signal CISPR 1 v CISPR 3 is applied, the output B1/IV pin 9 should be -9 V , and if the logic signal CISPR1 $\wedge \overline{\text { CISPR } 3}$ is applied, it should be +9 V .


### 5.3.1.3.3.2 Programmable Amplifiers B2/I, B2/II

- Switch on $B 6 / I$ and $B 6 / I V$ ( $A V, 20 \mathrm{~dB}$ ).
- Apply 2 V to ST1/17a. The voltage measured at the analog level output 7 a via $B 6 / I, B 2 / I, B 7 / I I I, B 3 / I I$ should be $3.55 \mathrm{~V}+2 \%$ and the voltage measured at $\mathrm{B} 2 / \mathrm{II}$ pin 15 should be $4 \mathrm{~V}+2 \%$ 。
- Apply a logic $L$ to the input (ST1 pin 2b).
- Measure voltage at the output to the sample/hold amplifier (ST1 pin 3a). It should be $2 \mathrm{~V}+2 \%$ 。
- When $B 6 /$ III is turned on ( $\log 40 \mathrm{~dB}$ or $\log 60 \mathrm{~dB}$ or etc.; see Table 4-1), the voltage at B2/II pin 15 should rise to $8 \mathrm{~V}+2 \%$ 。


### 5.3.1.3.3.3 Ad.justing the Logarithmic Converter

- Apply a variable $D C$ voltage to pin $17 a$ and switch on $B 6 / I$ by means of $B 7 / I$.
- Switch on LIN (B7/II).

Adjust with a digital voltmeter at pin 18a and measure at pin 7a.

- By alternately adjusting R37 and R40, the following requirements must be met:

$$
\begin{array}{lll}
\text { Pin 18a } \\
\left(\mathrm{v}_{\text {input }}\right)
\end{array} \quad \begin{array}{ll}
0.4 \mathrm{~V} \rightarrow \text { pin } 7 \mathrm{a} & 0.355 \mathrm{~V}+1 \% \\
\left.4.0 \mathrm{~V} \rightarrow \text { ( } \mathrm{V}_{\text {output }}\right) & 3.55 \mathrm{~V} \pm 1 \% .
\end{array}
$$

- Repeat these adjustments alternately until these values are reached.


### 5.3.1.3.3.4 Max. Level and Min. Level

For test setup see 5.3.1.3.3.3.
If the input voltage is less than 0.4 V , the MIN. LEVEL LED must light up (comparator output B3/III pin 9 approx. +9 V ).

If $\mathrm{V}_{\text {in }}>4 \mathrm{~V}$, the MAX. LEVEL IED must light up (comparator output B3/III pin 4 approx. +9 V ).

If these LEDs do not light up as required, check $T 3, T 4$ and the signal path up to the LEDS of the Display Unit.

It must be possible to scan the entire analog level indication LED array with the $V_{\text {in }}$ setting range 0.4 to 4 V . Beyond these range limits, the LEVEL MAX. and LEVEL MIN. LEED must light up.

### 5.3.1.3.4 Frequency Offset Signal Path

### 5.3.1.3.4.1 Offset Calibration

- After applying the supply voltage to the board, connect a voltage source with a control range of $+5 \mathrm{VEMF}\left(\mathrm{Z}_{\text {source }}=10 \mathrm{k} \Omega\right)$ to $\operatorname{ST1} / 14 \mathrm{~b}$. Set $\mathrm{V}_{\text {in }}$ to 0 V .
- Apply +5 V to $\mathrm{STl} / 5 \mathrm{a}$ (CAL. button).
- After this voltage has dropped ( $\mathrm{Stl} / 5 \mathrm{a}=\langle 0.8 \mathrm{~V}$ ), measure with a digital voltmeter at ST1/11a (2.5 V $+1 \%$ ).
- Press CAL. button 15 and then measure.
- Repeat this process until with $V_{\text {in }}(S T 1 / 14 b)=+5 V, 4 V+1 \%$ is obtained.
- Final testing: Set to the limit values and measure (after offset calibra* tion).

| $V_{\text {in }}$ | $\begin{aligned} & V_{\text {out }}(2 \%) \\ & \text { ST1/11a } \end{aligned}$ | $\begin{aligned} & V_{\text {out }}(2 \%) \\ & \text { STI } / 8 \mathrm{~b} \end{aligned}$ | $\begin{aligned} & V_{\text {out }}(T M T) \\ & \operatorname{STl} / 5 \mathrm{~b} \end{aligned}$ | LED <br> offset centre |
| :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} 0 \mathrm{~V}+0.025 \mathrm{~V} \\ -5 \mathrm{~V} \\ +5 \mathrm{~V} \end{gathered}$ | $\begin{aligned} & 2.5 \mathrm{~V} \\ & 1 \quad \mathrm{~V} \\ & 4.0 \mathrm{~V} \end{aligned}$ | $\begin{gathered} 0 \mathrm{~V} \\ -5 \mathrm{~V} \\ +5 \mathrm{~V} \end{gathered}$ | $\begin{array}{r} +5 \mathrm{~V} \\ <0.8 \mathrm{~V} \\ <0.8 \mathrm{~V} \end{array}$ | ON <br> OFF <br> OFF |

### 5.3.1.3.4.2 Peak-value Rectification

- Connect ST1/14b to ground and switch in B8/II ST1/4a=High (m).
- Connect digital voltmeter to B14/I pin 15 and set 2.5 V by means of R99.
- Connect digital voltmeter to B14/II pin 1 and set 2.5 V by means of R98.
- When a logic (TTL ) k is applied to 10 (peak value 50 ms ), the output voltage must drop to a maximum of $V_{\text {orig. }} / 2$ (storage capacitor).
- When logic $H$ is applied to select $m, \Delta f(-)$ at $S T 1 / 1 a$ and $m, \Delta f(+)$ at ST1/1b, no voltage difference must occur ( $<25 \mathrm{mV}$ ).
- The calibrated frequency offset signal (see 5.3.4.1) can be directly connected through to the output ST1/3a.
- The level can be taken to the peak-value rectifier via $B 8 / 3$ (AM measurement).


### 5.3.1.3.4.3 Frequency Correction Voltage

NOTE: Vary Rill only if the frequency of the mother oscillator is to be calibrated (pulled) for the synthesizers of the ESH 3 (the frequency responds only very slowly). If the frequency cannot be pulled, the setting range of Rlll at $\mathrm{STl} / 20 \mathrm{a}$, which should be +1 V to +9 V , can be checked using a digital voltmeter at ST1/20a.

Calibrate frequency:

- Select TWOPORT mode on the ESH 3 and connect a high-precision frequency counter to the generator output.
- Adjust by means of R111 the counter frequency to that set on the ESH 3.


### 5.3.1.3.4.4 AF Amplifier

- Apply supply voltage.
- Apply AF signal ( $\mathrm{V}_{\text {in }}=50 \mathrm{mV} / 1 \mathrm{kHz}$ ) to $\mathrm{SMl} / 18 \mathrm{~b}$.
- Connect an oscilloscope to STI/19a. Vout should be approximately $12 \mathrm{~V}_{\mathrm{pp}}$ if $\mathrm{SM} / 21 \mathrm{a}$ (volume control $50 \mathrm{k} \Omega$ ) is connected to ground and no loudspeaker is connected.
- Apply a logic $H$ to $S T 1 / 13 a(+5 \mathrm{~V})$ to switch off the AF amplifier (I7 is cut off).
- Measure gain Voltage gain $=$ approx. 33; $P_{\text {out }}=$ approx. 0.33 W into $16 \Omega$.
5.3.1.4 Synthesizer 2 (Y4)


### 5.3.1.4.1 Supply Voltages

| Pin | Voltage | Current |
| :---: | :---: | :---: |
| ST1, b24 | Ground | - |
| ST1, b23 | $+5.25 \mathrm{~V}+0.1 \mathrm{~V}$ | approx. 110 mA |
| ST1, 222 | $+10 \mathrm{~V}+0.1 \mathrm{~V}$ | approx. 25 mA |
| ST1, a23 | $+12 \mathrm{~V}+0.5 \mathrm{~V}$ | approx. 60 mA |
| ST1, a24 | $+25 \mathrm{~V}+0.2 \mathrm{~V}$ | approx. 6 mA |
| ST1 , be2 | $-10 \mathrm{~V}+0.1 \mathrm{~V}$ | approx. 3 mA |
| ST1, ab19 | +1 to +9 V | approx. <1 mA |

### 5.3.1.4.2 Reference

- Remove ST5. Connect spectrum analyzer to ST3. $f=60 \mathrm{MHz}$, level: approx. -20 dBm .
- Adjust I68 so that the oscillator starts oscillating and the frequency is 60 MHz if +5 V is present at ST1.ab19.
- Set maximum level by means of C 160 .
- Vary voltage at STI, ab19 between +1 and 9 V at the same time checking the spectral purity of the signal.
- Check the TTL reference signals at

| ST1.b18 | $:$ | $f=100 \mathrm{kHz}$ |
| :--- | :--- | :--- |
| ST1.ab1 | $:$ | $f=500 \mathrm{~Hz}$ |
| B41,1 | $:$ | $f=500 \mathrm{~Hz}$ |
| B41,3,13 | $:$ | $f=1 \mathrm{kHz}$ |

- Apply external reference signal to $\operatorname{STH}(f=5 / 10 \mathrm{MHz}$ ) from $1-V, 50-\Omega$ source. Frequency error better than $2 \times 10^{-6}$.
- Apply +5 V comesponding to external reference to ST1.a2 and b2.
- Measure frequency at ST3, the accuracy of which must be equal to that of the external reference.

This frequency must not change when the voltage at ST1.abl9 is varied between +1 and +9 V .

- Connect the spectrum analyzer to BU1 pin 8 and ground, which permits detailed display of the spectral purity of the reference oscillator.
- Unwanted oscillations of T1 and TY must be avoided by short circuiting the coils I4 and L21, respectively.

Discrete non-harmonic spurious signals should be down $>80 \mathrm{~dB}$.
Sideband noise 10 kHz away should be down $>140 \mathrm{~dB} / \mathrm{Hz}$.

### 5.3.1.4.3 $\quad 50-$ to $-51-\mathrm{MHz}$ Oscillator

- Connect spectrum analyzer to ST5 R116/ground.
- Remove shortcircuit across Lil.
- Adjust frequency to 50.05 MHz , connecting $\$ T 1 . a 3$ to 2.14 as follows:
$\mathrm{H}=\mathrm{a3}$, a 5
$L=a 4, a 6$ to a14.
- Set C42 to mid-position.

(Adjustment by means of R117)
Check at ST1.b13:
$H=$ oscillator is synchronized
$\mathrm{L}=$ oscillator is not synchronized.

If oscillator not synchronized, then use a scope to check the logic levels or oscillograns at the points given, by reference to the circuit diagram:
B19, pin 11
B20, pin 8
B22, pin 6
B24, pins $1,3,10,13,4,12$
B25, pin 12
B18, pin 23
B16, pin $14,1 \mathrm{kHz}$ if synchronized
B16, pins 13. 12
B17, pins 11, 8, 9, 2

- Check voltages at B15:

Pin 3: $+2.5 \mathrm{~V}+0.2 \mathrm{~V}$
Pin 7: +25 V $\ddagger 1 \mathrm{~V}$
(Adjust by means of C24:)
Pin 6: $+6.5 \mathrm{~V} \pm 0.3 \mathrm{~V}$.

- Apply required logic levels to vary the frequency between 50.000 and
50.999 MHz and watch at the same time the synchronization indication (b13).
- Check the spectral purity:

Discrete non-harmonic spurious signals ................... down $>80 \mathrm{~dB}$
Sideband noise 10 kHz away ................................ $>140 \mathrm{~dB} / \mathrm{Hz}$
Level at STB: $f=65.0000$ to $65.0999 \mathrm{MHz} \ldots \ldots \ldots . . \begin{aligned} & -16 \mathrm{dBm}+2 \mathrm{~dB}\end{aligned}$


Fig. 5-1 Typical tuning characteristic of the 50-to-51aMHz oscillator
2.3.1.4.4 $66-\mathrm{MHz}$ Oscillator

- Connect spectrum analyzer to ST2.
- Remove short-circuit across IH.
- Rotate core of I4 inwards until the oscillator starts to operate. Adjust C13 for maximum.

Frequency: corresponds to frequency of Q 1.
- Check synchronization indication STI.bl4:
$H=$ oscillator synchronized
$I_{1}=$ oscillator not synchronized.

If oscillator not synchronized, then use a scope to check the logic levels or oscillograms at the points given, by reference to the circuit diagram:

B1, pin 11
B2, pin 8
B3, pin 6
B10, pins 1, 3, 10, 13, 4, 12
B5, pin 23
B7, pins 14, 3, 1, 2, 13
B8, pins $10,5,6$

- Check voltages at B6:

$$
\begin{aligned}
& \text { pin 3:.......................... } \mathrm{V}+0.2 \mathrm{~V} \\
& \text { pin 7: ....................... }+25 \mathrm{~V}+1 \mathrm{~V} \\
& \text { pin 6: ....................... see illustration }
\end{aligned}
$$

- Apply a logic $H$ to $S T 1 . b 3$ : The frequency is increased by 1.5 kHz .
- Apply a logic $H$ to STI.b4: The frequency is decreased by 1.5 kHz .
- Check the spectral purity:

Discrete non-harmonic spurious signals ............. down $>80 \mathrm{~dB}$
Sideband noise 10 kHz away ............................................. $>140 \mathrm{~dB} / \mathrm{Hz}$


Fig. 5-2 Typical tuning characteristic of the 2nd oscillator ( 66 MHz )

### 5.3.1.5.1 Supply Voltages

| Pin | Voltage | Current | Frequency |
| :---: | :---: | :---: | :---: |
| ST1, b24 | Ground | - | - |
| ST1, b23 | $+5.25+0.1 \mathrm{~V}$ | approx. 105 mA | 0 |
| ST1, a22 | +10 +0.1 V | approx. 35 mA |  |
| ST1, a23 | $+12+0.2 \mathrm{~V}$ | max. 75 mA | 0 |
| STI, a24 | $+25+0.1 \mathrm{~V}$ | approx. 4 mA | 0 |
| ST1, b22 | $-10+0.1 \mathrm{~V}$ | approx. 6.5 mA | 0 |
| St1, b18 | THT | $<1 \mathrm{~mA}$ | 100 kHz |
| ST1, a7 | TIT | $\cdots \quad<1 \mathrm{~mA}$ | - |
| b8 | $100-\mathrm{kHz}$ decade |  | - |
| $\begin{aligned} & \text { a9 } \\ & \text { b10 } \end{aligned}$ |  |  | - |
| all | TTL | $<1 \mathrm{~mA}$ | - |
| 212 |  |  | $\infty$ |
| b13 | 1 MMHz decade |  | $=$ |
| al4 |  |  |  |
| al6 | TIT | $<1 \mathrm{~mA}$ | $\cdots$ |
| a15 | $10-\mathrm{MHz}$ decade |  | - |
| ST2 | $-17 \mathrm{dBm}+1 \mathrm{~dB}$ | - | 65.0 to 65.1 MHz |

### 5.3.1.5.2 Adjustment of 75/85/95/105-MHz Osci11ators

- Connect frequency counter (50 $\Omega$ ) to $\mathrm{ST3}$.
- Remove B3 and connect R15 to +10 V .

The oscillators are controlled by the TIT levels at the input lines of the decades depending on the receive frequency, and pretuned by a $D / A-$ converted voltage.

- Check the pretuning voltage at $S T 1$, al7 and $R 16$ :

|  | ST1, a17 | R16 |
| :--- | :--- | :--- |
| All decade inputs at logic I | $0 \pm 5 \mathrm{mV}$ | $3.8 \mathrm{~V}+20 \mathrm{mV}$ |
| All inputs of the $100-\mathrm{kHz}$ <br> and 1 a MHz decades at 10 gic H | $7.4 \mathrm{~V}+0.2 \mathrm{~V}$ | $17.8 \mathrm{~V}+0.8 \mathrm{~V}$ |

- The tuning range is adjusted by means of the variable capacitor, and the initial frequency is adjusted by means of the variable inductance (see Fig. 5-7 in the appendix).

| Oscillator | Decade driving signal | Adjust with |
| :---: | :---: | :---: |
| 75 to 85 MHz | 0.000 to 9.9 MHz | C3, Li2 |
| 85 to 95 MHz | 10.0 to 19.9 MHz | C13. L13 |
| 95 to 104.0 MHz | 20.0 to 29.9 MHz | C23. L1 4 |

## 5.3 .1 .5 .3 Adjustment of Output Level

- Connect spectrum analyzer (50 $\Omega$ ) to ST3.
- Switch the decades in $1-\mathrm{MHz}$ steps.
- Adjust the output level by means of C38 over
the frequency range from 75 to 105 Mifz to $\ldots \ldots . . \ldots .{ }^{2}+23 \mathrm{dBm}+2 \mathrm{~dB}$


## 5.3 .1 .5 .4 Adjustment of $65-\mathrm{MHz}$ Amplifier

Connect sweep generator (50 $\Omega$ ) to ST2 and apply
level of $-17 \mathrm{dEm}$

Frequency range 60 to 70 MHz

Connect display section of the sweep generator to $S T 7$. Make coarse adjustment of L31, L32, L33 for masimum at 65.5 MHz . Connect sweep generator to ST7 (via capacitor). Level and frequency same as above. Connect display section to ST6. Make coarse adjustment of L36, L37, L33 for maximum at 65.5 MHz . Connect sweep generator to ST2. Make final adjustments to reach selectivity curve shown below.


### 5.3.1.5.5 Adjustment of 10 to $-40-\mathrm{MHz}$ Amplifier

Apply signal between 65 and 65.1 MHz with a level of $16+1 \mathrm{dBm}$ to $\mathrm{ST2}$. Check level at ST5 with the aid of a spectrum analyzer while varying the oscillator frequency between 75 and 105 MHz Required level
$-21 d B m+2 d B$
Frequency ................................................... 10 to 40 MHz
Check level at ST4 with the aid of a spectrum analyzer by varying the oscillator frequency between 75 and 105 MHz Required level $+2 d B m+3 d B$

### 5.3.1.5.6 Checking the Synchronization Indication

Synchronization indication .............................. Iogic L
(B3 removed)

### 5.3.1.5.7 Checking the Synchronization

## Insert B3.

Apply 100 kHz reference (crystal-controlled, with low sideband noise).
Synchronization indication .............................. logic Hi
Voltage at R15 (depending on f) ..................... $>+3$ to $<+19 \mathrm{~V}$

### 5.3.1.5.8 Checking the Sideband Noise

The signal at ST2 should have minimal sideband noise.
1 kHz away ........................................... down typ. $90 \mathrm{~dB} / \mathrm{Hz}$
10 kHz away .......................................... down typ. $135 \mathrm{~dB} / \mathrm{Hz}$
The $100-\mathrm{kHz}$ reference must be crystal-controlled with low sideband noise.
The typical sideband noise characteristic is shown in Fig. 5-3.
Excessive variations at A can be corrected by varying the gain of B3 (R95). If there are considerable, abrupt frequency changes or warm-up transients, then that points to problems with the oscillators.

Required rejection of discrete signals over the range 0 to 50 kFlz from the carrier (measured at an analyzer IF bandwidth of
10 Hz ) .................................................................... $>100 \mathrm{~dB}$
Time required for synchronization after any change
in frequency ............................................... $<100 \mathrm{~ms}$


Fig. 5-3 Characteristic of typical sideband noise
5.3.1.6 Filter Control (Y6)
5.3.1.6.1 Supply Voltages

| Pin | Voltage |  | Current | Frequency |
| :--- | :--- | :--- | :--- | :--- |
| ST1.b24 | Ground |  | - | - |
| ST1.b23 | +5.25 V | $\pm 0.1 \mathrm{~V}$ | approx. | 3 mA |
| ST1.a22 | +10 V | +0.1 V | approx. 1.5 mA | 0 |
| ST1.a24 | +30 V | +0.5 V | approx. 0.7 mA | 0 |
| ST1.b22 | -10 V | $\pm 0.1 \mathrm{~V}$ | approx. 3.5 mA | 0 |
| ST1.b17 | 0 to 7.5 V |  | $<$ | 1 mA |
| ST1.ab1 | TIT |  | 0 | 0 |

### 5.3.1.6.2 Checking the Digital Section

Connect ST1.b20 via $470 \Omega$ to ground.
Apply logic levels to pins ST1, 3 to 16.
Truth table

| $f(\mathrm{MHz})$ | Range | High level Sti, a... | STI, a2 |
| :--- | :--- | :--- | :--- |
| 0 to 0.14 | 1 | 21 | L |
| 0.15 to 0.19 | 2 | 19 | L |
| 0.2 to 0.27 | 3 | 17 | L |
| 0.28 to 0.38 | 4 | 15 | L |
| 0.39 to 0.53 | 5 | 13 | L |
| 0.54 to 0.74 | 6 | 11 | L |
| 0.75 to 1.04 | 7 | 9 | L |
| 1.05 to 1.44 | 8 | 7 | H |
| 1.45 to 1.99 | 9 | 6 | H |
| 2.0 to 2.69 | 10 | 8 | H |
| 2.7 to 3.69 | 11 | 10 | H |
| 3.7 to 5.19 | 12 | 12 | H |
| 5.2 to 7.19 | 13 | 14 | H |
| 7.2 to 9.99 | 14 | 16 | H |
| 10 to 19.99 | 15 | 18 |  |
| 20 to 29.99 | 16 | 20 |  |

While the range is being selected, $S t 1 . b 20$ is at low level. Maximum time required for range selection $\ldots . . . .$.

### 5.3.1.6.3 Checking the Analog Section (Pre-adjustment)

Increase the input voltage at $\mathrm{ST1.b17}$ in steps of 75 mV from OV to +7.4 V . The resulting output voltage at $\mathrm{STI.24}$ and the trimmers used are shown in Table 5-1.

## Table 5-1

| Ingut voltage (V) | 0 | +0.75 | +3.75 | +6.75 | +7.4 |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Adjusting element <br> Output voltage (V) | $R 79$ <br> $+3+0.1$ | $R 55$ <br> $+6+0.1$ | $R 70$ <br> $+13.75+0.1$ | $R 89$ <br> $23+0.1$ | $R 103$ <br> $26+0.1$ |

Make final adjustment together with Filter 2 (Y8) for minimum input reflection coefficient in range 15.

### 5.3.1.7 Filter 1 (Y7)

Supply voltages:

| Pin | Voltage | Current |
| :--- | :--- | :--- |
| ST1.b24 | Ground | - |
| ST1.b23 | $+5.25 \mathrm{~V} \pm 0.1 \mathrm{~V}$ | approx. 30 mA |

Switch in the desired filter by applying a logic $H$ to pins STl.a7 to a21. Adjust filters 1 to 8 for a reflection coefficient characteristic in accordance with Fig. 5-4 using a sweep-frequency network analyzer.


Fig. 5-4 Reflection coefficient characteristic of the fixed-tuned filters
Reflection coefficient within the filter range ........... $<10 \%$
Reflection coefficient at the cut-off frequencies ........< $13 \%$
Insertion loss .................................................... $<0.5 \mathrm{~dB}$

Test setup:


Table 5-2

| Filter | $\begin{aligned} & \text { Cut-off } \\ & \mathrm{f}_{1} \end{aligned}$ | equencies $f_{u}$ | Adjust with | Logic ${ }^{\text {H }}$ to STI.a... |
| :---: | :---: | :---: | :---: | :---: |
| 1 | $10^{*} \mathrm{kHz}$ | 150 kHz | 12, 13, I4 | 21 |
| 2 | 150 kHz | 200 kHz | L5, I6, L7 | 19 |
| 3 | 200 kHz | 280 kHz | L10, L11, Li2 | 17 |
| 4 | 280 kHz | 390 kHz | L15, L16, L17 | 15 |
| 5 | 390 kHz | 540 kHz | L20, L21, L22 | 13 |
| 6 | 540 kHz | 750 kHz | L25, I26, L27 | 11 |
| 7 | 750 kHz | 1.05 MHz | L30, L31, L32 | 9 |
| 8 | 1.05 MHz | 1.45 MHz | L35, L36, L37 | 7 |

$f_{1}=20 \mathrm{~Hz}$ for model 56 (subassembly model 53 )
5.3.1.8 Filter 2 (Y8)
5.3.1.8.1 Supply Voltages

| Pin | Voltage | Current |
| :--- | :--- | :--- |
| ST6.b24 | Ground | - |
| ST6.b23 | $+5.25 \mathrm{~V}+0.1 \mathrm{~V}$ | approx. $30 / 60 \mathrm{~mA}$ |
| ST6.a2 | Logic H | $<0.5 \mathrm{~mA}$ |
| ST6.a4 | +3 to +28 V | $<0.1 \mathrm{~mA}$ |

Switch in desired filter by applying a logic $H$ to pins ST6.a6, 8 to 20. Adjust filters 9 to 14 similarly to 5.3 .1 .7 in compliance with the following table:

Table 5-3

| Filter | Cut,off frequencies <br> $f_{1}$ |  | Adjust with | Logic H to ST.a... |
| :--- | :--- | :--- | :--- | :--- |
| 9 | 1.45 MHz | 2.0 MHz | I40, I41, I42 | 6 |
| 10 | 2.0 MHz | 2.7 MHz | L45, I46, I47 | 8 |
| 11 | 2.7 | MHz | 3.7 MHz | L50, L51, L52 |
| 12 | 3.7 MHz | 5.2 MHz | L55, L56, L57 | 12 |
| 13 | 5.2 MHz | 7.2 MHz | I60, I61, I62 | 14 |
| 14 | 7.2 MHz | 10.0 MHz | I65, I66, I67 | 16 |

Feed in test signal to ST7 (terminate with $50 \Omega$ at SI8).

### 5.3.1.8.2 Adjustment of Filter 15

Adjust L72, L73, C156 as follows:
$f=11 \mathrm{MHz}, \mathrm{V}$ into $\mathrm{ST} .24=+6 \mathrm{~V}+0.2 \mathrm{~V}$
$f=15 \mathrm{MHz}, \mathrm{V}$ into $5 T 6.2 .4=+13.75 \mathrm{~V} \pm 0.3 \mathrm{~V} \quad \mathrm{r}<10 \%$
$\mathrm{f}=19 \mathrm{MHz}, \mathrm{V}$ into $\mathrm{SI} .24=+26 \mathrm{~V}+0.3 \mathrm{~V}$
$6-\mathrm{dB}$ bandwidth of filters 15 and 16 ............................. $<6 \mathrm{MHz}$
Attenuation at $f_{u} / 2$................................................... 18 dB

### 5.3.1.8.3 Ad.justment of Filter 16

Adjust L78, L79, C162, C164 for minimum reflection coefficient of the ESH 3 over the frequency range 20 MHz to 29.9999 MHz after having adjusted filter 15 using Y6.

C164: maximum effect at $f_{1}$
C162: maximum effect at $f_{u}$.
After adjustment, fix the cores of L72, L73, L78 and L79 with the cable tie provided for this purpose and secure the coils using any type of varnish or glue that is suitable for RF applications.

### 5.3.1.9 Mixers 1 and 2 (Y9)

Note: The pin designations in brackets (e.g. (x7)) or the component designations in brackets (e.g. (V111)) refer to model 56 of the ESH 3. Unless otherwise stated, the adjustment procedure is identical.

### 5.3.1.2.1 Supply Voltages

| PIn | Voltage | Current |
| :--- | :--- | :--- |
| ST1.b24 | Ground | - |
| ST1.a22 | $+10+0.1 \mathrm{~V}$ | approx. 10 mA |
| ST1.a23 | $+12+0.3 \mathrm{~V}$ | approx. 60 mA |
| ST1.a.24 | $+25+0.1 \mathrm{~V}$ | approx. 25 mA |
| ST1.b22 | $-10 \div 0.1 \mathrm{~V}$ | approx. 5 mA |

### 2.3.1.9.2 Adjustment of Input Low-pass Filter

- Connect network analyzer to ST2.
- Remove BU1.
- Connect 50- $\Omega$ termination between ST8 pins 1 and 4 ( $r<2 \%$ ). Tune L1, L2, L3, L4, L5 for reflection coefficient .......... $<10 \%$ (over frequency range 0 to 30 MHz ).
5.3.1.9.3 Adjustment of $75-\mathrm{MHz}$ Amplifier
- Using R1 set operating point of T ( V 111 ) to .............. $22 \mathrm{~mA} \pm 2 \mathrm{~mA}$
- Feed frequency of 75 MHz into $\operatorname{ST9} / 2,3$ ( $\mathrm{X} 9 / 1,2$ ) using a sweep generator.
- Connect sweep display unit to ST10/1,4 (X10/A1,2)
- Adjust C14 for maximum at $f=75 \mathrm{MHz}$.

- Connect sweep display unit to ST14 (X14).
- Adjust amplifier T2, T14 and T15 (V200, V140 andV150).
- Adjust C117 for maximum at $f=75 \mathrm{MHz}$.
- Required gain .......................................................................... $\mathrm{dB} \pm 3 \mathrm{~dB}$
- (The ESH 3 is supplied with the $75-\mathrm{MHz}$ amplifier permanently switched on (ST10 (X13) plugged on).


### 5.3.1.9.4 Adjustment of Diplexer of 2nd Mixer

- Connect sweep generator toST11.2 (X11/B) and ST12.2 (X12) (50-ת system) and adjust L 18 for maximum attenuation at 9 MHz .
- Connect sweep generator to ST11. 2 ( $\mathrm{X} 11 / \mathrm{B}$ ) and. ST7 (X7) ( $50-\bar{\Omega}$ system) and adjust C31 for maximum gain at 9 MHz .


### 5.3.1.9.5 Adjustment of $9-\mathrm{MHz}$ Amplifier

Connect a sweep generator to ST11 (2,3) (X11/B) and ST7 (X7) (50-ת system) for the adjustment.

## $5.3 \cdot 1 \cdot 9.6 \quad 500-\mathrm{Hz}$ Bandwidth

- Connect ST1.a3 (X1.A3) to ground.
- Adjust R63 for maximum gain.
- Adjust C53 for minimum ripple.
- Note gain.


### 5.3.1.9.7 2.4-kHz Bandwidth

- Connect ST1.a5 (X1.A5) to ground.
- Set R69 to mid-position.
- Adjust C64, C67 for minimum ripple.
- Adjust R69 for same gain as with $500-\mathrm{Hz}$ bandwidth.


### 5.3.1.9.8 $10-\mathrm{kHz}$ Bandwidth

- Adjust R 76 for same gain as with $500-\mathrm{Hz}$ and with $2.4-\mathrm{kHz}$ bandwidth. Gain of $9-\mathrm{MHz}$ amplifier .................................................. $27 \pm 1 \mathrm{~dB}$


### 5.3.1.9.9 Adjustment of Overload Detection

- Insert BU3 and BU4. (Connect X10A to X10B and X11A to X11B) The 2nd oscillator ( $\mathrm{f}=66.00 \mathrm{MHz}$, level $=+7 \mathrm{dBm} \pm 1 \mathrm{~dB}$ ) is connected to ST5 (X5).


### 5.3.1.9.9.1 Overload Detection 2

Feed $75.00-\mathrm{MHz}$ signal to ST9 pins 2 and 3 (X9/B).
At a level of $-22 \mathrm{dBm} \pm 2 \mathrm{~dB}$, the logic signal at $\mathrm{ST} 1 . a 7$ (X1.A7) should change from H to L. Adjust with R35 (increasing R35 raises the response threshold).

### 5.3.1.9.9.2 Overload Detection 1

Feed a signal of frequency 75 to 135 MHz into $\operatorname{ST9}$ pins 2 and 3 (X9/B).
At a level of $\pm 3 \mathrm{dBm} \pm 1 \mathrm{~dB}$, the logic H at ST1.a9 (X1.A9) should change to logic L. Adjust with R18 (increasing R18 raises the response threshold).

### 5.3.1.9.9.3 Checking the Level at ST4 (X4)

Feed $+23 \mathrm{dBm} \pm 1 \mathrm{~dB}$ (frequency range 75 to 105 MHz ) into ST 3 .
Output level réquired at $\operatorname{ST} 4$ ( 84 ) .......................................... $-15 \mathrm{dBm} \pm 2 \mathrm{~dB}$ (Adjust with R94 (R208)).

### 5.3.1.9.9.4 Adjustment of the 1st Mixer

(to be carried out on model 56 of ESH 3 only).

- Connect X8.A1 to X8.B2 and X9.A1 to X9.B2.

Terminate X 2 with $50 \Omega$.

- Feed 1st oscillator (level to $+23 \pm 2 \mathrm{dBm}$, frequency 75 to 105 MHz ) at X 3 .
- Connect spectrum analyzer to X14.
- Level measured at X14 at $75-\mathrm{MHz}$ oscillator frequency should be <-32 dBm (Note effect of cover on soldered side).
- An adjustment is possible by bending the terminal wires 3 and 6 of transformers T4 and T6.
- Following the adjustment, pins of T4 must be affixed using yellow UHU adhesige.
- Recheck level at X14.
5.3.1.9.9.5 Checking the Level at ST6

Feed $+7 \mathrm{dBm} \pm 1 \mathrm{~dB}$ (frequency 66 MHz ) into ST5.
Output level required at ST6 ............................................ $-15 \mathrm{dBm} \pm 2 \mathrm{~dB}$
(Adjust with R96).
Check the following performance specifications of Y9 using test setups as in section 3.2 :

```
Input reflection coefficient 10 kHz to 30 MHz ........ < 20%
Noise figure 100 kHz to 30 MHz ........< < 13 dB
Oscillator reradiation }\quad75\textrm{MHz}\mathrm{ to 105 MHz ....... < 20 dBurV
Image frequency rejection 150 MHz to 180 MHz ....... > 90 dB
IF rejection }75\textrm{MHz}.......>100 d
IM rejection a }\mp@subsup{\textrm{d}}{2}{}-3rd order intercept point ......... > +20 dBm
(see specifications)
IF bandwidths }10\textrm{kHz},2.4\textrm{kHz},500\textrm{Hz
(see specifications)
Gain ..................................................................... }35 +1 d
```

5.3.1.10 Calibration Generator (YiO)
5.3.1.10.1 Supply Voltages

| Pin | Voltage/Level | Current | Frequency |
| :--- | :--- | :--- | :--- |
| ST1.b24 | Ground | - | - |
| ST1.b23 | $+5.25 \pm 0.1 \mathrm{~V}$ | approx. $0 / 65 \mathrm{~mA}$ | 0 |
| ST1.a22 | $+10 \pm 0.1 \mathrm{~V}$ | approx. $65 / 6 \mathrm{~mA}$ | 0 |
| ST1.b22 | $-10 \pm 0.1 \mathrm{~V}$ | approx. 3 mA | 0 |
| ST1.al | ITL |  | 500 Hz |
| ST4 | -15 dBn | - | 30 kHz |
| ST5 | -13 dBm | - | 8.97 MHz |
| ST6 | -15 dBm | - | $66 \mathrm{MHz}(+1.5 \mathrm{kHz)}$ |
| ST3 | -15 dBm | - | 75 MHz to 105 MHz |

5.3.1.10.2 Adjustment of Sinewave Generator

- Connect +5 V to STI.a4.
- Connect spectrum analyzer to ST2.
- Connect voltmeter to ST1.a15.
- Adjust for maximum level on the spectrum analyzer and minimum voltage on the voltmeter at:

9 MHz by means of C 9 ( $\mathrm{ClO}, \mathrm{C} 20$ )
75 MHz by means of C45, C52, C61, C66.
The adjustment of C61, C66 is interactive.

Level at $S T 2$ in the frequency range 10 kHz to $30 \mathrm{MHz} \ldots 80 \mathrm{~dB} \mathrm{\mu V}+3 \mathrm{~dB}$ (Variable with R84; measured with a power meter.)

Spectral purity of the output signal at ST2:
Non-harmonic spurious signals ................................. $>40 \mathrm{~dB}$ down
Check the control voltage at ST1.a15:
When tuning through the range from 10 kHz to 30 MHz , the voltage must not change abruptly and there must be no AC voltage superimposed on it.

Check by means of oscilloscope; typically .................... 0.8 V to 4 V

### 5.3.1.10.3 Ad,justment of Temperature Compensation of Sinewave Generator

 The temperature compensation can be adjusted in several cold/warm cycles. For the relationship between change in level and required sense of rotation of R77 see Table 5-4.
## Table 5-4

| Increase in temperature causes level increase. <br> Decrease in temperature causes level drop. | Turn R77 <br> counterclockwise. |
| :--- | :--- |
| Increase in temperature causes level drop. | Turn R77 |
| Decrease in temperature causes level increase. | clockwise. |

Permissible total change in level due to change in frequency and temperature over the range from
-10 to $+45^{\circ} \mathrm{C}$ $<0.5 \mathrm{~dB}$

Required level at $S T 2$......................................... $80 \mathrm{~dB} \mu \mathrm{VV}+0.3 \mathrm{~dB}$ $=-27 \mathrm{dBm}+0.3 \mathrm{~dB}$
over the total temperature and frequency range.

## Table 5-5

Input level (measured with spectrum analyzer)

| Pin | Level | Frequency |
| :--- | :--- | :--- |
| 4 | $-15 \mathrm{dBm} \pm 1 \mathrm{~dB}$ | 30 kHz |
| 5 | $-13 \mathrm{dBm} \pm 1 \mathrm{~dB}$ | 8.97 MHz |
| 6 | $-15 \mathrm{dBm} \pm 1 \mathrm{~dB}$ | 66 MHz |
| 3 | $-15 \mathrm{dBm} \pm 1 \mathrm{~dB}$ | 75 MHz to 105 MHz |

Table 5-6
Mixer level
(measured by means of RF probe at pin 8)

| Mixer | Level | Frequency |
| :--- | :--- | :--- |
| B 1 | $0 \mathrm{dBm}+2 \mathrm{~dB}$ | 8.97 MHz |
| B 3 | $0 \mathrm{dBn}+2 \mathrm{~dB}$ | 66 MHz |
| B5 | $0 \mathrm{dBm}+2 \mathrm{~dB}$ | 75 MHz to 105 MHz |

Level required at C7O ( $f=10 \mathrm{kHz}$ to 30 MHz , with AGC, measured by means of high-impedance probe) ............... $5 \mathrm{mV}+3 \mathrm{~dB}$
Level at ST7
$-67 \mathrm{dBm}+0.3 \mathrm{~dB}$
Level variation by Rllo $+0.7 \mathrm{~dB}$

### 5.3.1.10.4 Checking the CISPR 3 or CISPR 1 Pulse Generator

- Disconnect STI.a4 from +5 V , and instead connect either STI.a6 (CISPR 3) or 25 (CISPR 1) to +5 V .
- Check pulse at ST7, terminated with $50 \Omega$, by means of a $1-\mathrm{GHz}$ oscilloscope:


## Table 5-7

|  | Logic H <br> to ST1 | PRF | Pulse energy <br> $\mu$ Vs (EMF) <br> $+10 \%$ | Adjust <br> with |
| :--- | :--- | :---: | :--- | :--- |
| CISPR 3 | 26 | 25 Hz | 1.35 <br> CISPR i | 25 | | 100 Hz |
| :--- | $0.0316 \quad$| R89 |
| :--- |
| R91 |



Rise time $t_{r}=3 \mathrm{~ns}$
a) with CISPR 1 pulse


Rise time $t_{r}=4 \mathrm{~ns}$
b) with CISPR 3 pulse

## Fig. 5-5 Pulse calibration

For a rough measurement, the pulse energy can be graphically integrated or calculated according to the following simplification:


$$
\begin{aligned}
& \text { Pulse area }=V \cdot t_{50 \%} \\
& \text { Accordingly, with CISPR } 1 \text { pulse (Fig. } 5-5 \mathrm{a} \text { ), } \\
& \text { the pulse area }=2.8(\mathrm{~V}) \cdot 5.6(\mathrm{~ns})=0.01568 \mu \mathrm{Vs}
\end{aligned}
$$

corresponding to an EMF pulse energy of $0.03136 \mu \mathrm{Vs}$.
This calibration pulse of the ESH 3 is 20 dB below the CISPR 1 standard pulse.

Fig. 5-5b: CISPR 3 pulse
$\mathrm{V} \cdot \mathrm{t}_{50 \%}=3(\mathrm{~V}) \cdot 210(\mathrm{~ns})=0.63 \mu \mathrm{Vs}$
$V \cdot t_{50 \%}=1.26 \mu \mathrm{Vs}$ (EMF pulse area)
Hence, this calibration pulse is 20 dB below the CISPR 3 standard pulse ( $13.5 \mu \mathrm{Vs}$ ).
5.3.1.11 Mixer 3 (Y11)
5.3.1.11.1 Supply Voltages

| Pin | Voltage | Current | Frequency |
| :--- | :--- | :--- | :--- |
| ST1.b24 | Ground | - | - |
| ST1.b23 | $-5.25+0.1 \mathrm{~V}$ | approx. 12 mA | 0 |
| ST1.a22 | $+10 \mathrm{~V}+0.1 \mathrm{~V}$ | approx. 48 mA | 0 |
| ST1.b22 | $-10 \mathrm{~V}+0.1 \mathrm{~V}$ | approx. 18 mA | 0 |
| ST1.b1 | TIL | - | 500 Hz |

### 5.3.1.11.2 Ad,justment of 8.97 mHz Oscillator

- Apply exact $500-\mathrm{Hz}$ reference to $\mathrm{ST1.b1}$ (TTL level squarewave; for repair purposes, use ESH 3 reference)
- Measure control voltage $\mathrm{V}_{\text {control }}$ at $\mathrm{R} 40 / \mathrm{R} 41$
Adjust $C 43$ so that $V_{\text {control }}$
$5 \mathrm{~V} \pm 1 \mathrm{~V}$
(If adjustment is not possible, vary Lil slightly)
- Measure output level on $S T 3$................................... $13 \mathrm{dBm}+2 \mathrm{~dB}$
- Measure oscillator frequency at STJ ....................... 8.97 MHz
(The accuracy of the measured value depends on the accuracy of the reference used.)
- Check synchronization indication at STl.b2: logic H (loop is synchronized).


### 5.3.1.11.3 Ad,justment of Basic Gain between ST2 and ST5

a) Initial settings:

- Settings on the ESH 3:

IF bandwidth 5: 500 Hz to $10 \mathrm{kHz}(S T 1 . a 11=\mathrm{L})$
IF attenuation 40, 41: 40 dB

- Connect ST1. bl 12 to +10 V .
- Apply external $D C$ voltage of -5 to +10 V to $\mathrm{STl} . a 12$.
- Set external DC voltage to $\geqq 4 \mathrm{~V}$ (= maximum gain).
$V_{E M F}$ at ST2 with $f=9.000 \mathrm{MHz}=20 \mathrm{mV}$.
Measure at ST5 via a high impedance using a millivoltmeter.


## Test setup:

## $90 . \mathrm{MHz}$


b) Adjustment

Set output voltage $V_{\text {out }}$ by means of R20.


### 5.3.1.11.4 Checking the Noise Filter Bandwidth

Measure between ST2 and STS with any IF bandwidth other than 0.2 kHz (ST1.a11 = L). Vary the signal-generator frequency to determine the upper and lower $1-\mathrm{dB}$ and $3-\mathrm{dB}$ cut-off frequencies of the noise filter.

Table 5-8

|  | $f_{\text {cut-off }}$ lower $/ \mathrm{kHz}$ | $f_{\text {cut-off upper }} / \mathrm{kHz}$ |
| :--- | :--- | :--- |
| $1-\mathrm{dB}$ cut-off frequencies | $25 \pm 1$ | $35 \pm 1$ |
| $3-\mathrm{dB}$ cut-off frequencies | $23 \pm 1$ | $37 \pm 1$ |

The pass-band characteristic should be approximately flat and must not exhibit a dip.

### 5.3.1.11.5 Checking the Calibration Gain Control Range

- Apply +10 V to ST1.bl2 (calibration $=0 \mathrm{~N}$ ).
- Feed DC voltage $V_{\text {control }}$ into ST1.al2.
- Vary the voltage at the base of Tl from 0 to +5 V .
－The gain between ST2 and ST5 should be variable between about -15 dB and 0 dB referred to maximum gain $G_{\max }$ ．

When the voltage at the base of Tl is $2 \mathrm{~V}_{\text {s }}$ the gain is typically 6 dB below $G_{\text {max }}$ ．

## 5．3．1．11．6 Checking the IF Attenuation Steps

Test setup：


Use as accurate an attenuator as possible，such as the DPVP．
－Set $V_{\text {control }} \mathrm{CAL}$ at the base of Tl again to about 2.0 V ．
－Set the attenuator to 50 dB 。
－Set the IF attenuation of the 3rd mixer to 0 dB （＝maximum gain）．
－Adjust signal generator level such that the milliwoltmeter gives a deflection approximately $2 \mathrm{~dB}<100 \mathrm{mV}$（near the $0-\mathrm{dB}$ marker）．
－Increase the IF attenuation of the 3 rd mixer in steps of 10 dB and reduce the attenuator setting in steps of 10 dB 。
－Observe deflection on milivoltmeter：The difference（ $=$ attenuation error） should be less than 0.1 dB ．The sum of the attenuation errors should be approximately 0 dB 。

## 5．3．1．11．7 Checking the IF Output Amplifier

Use the same test setup as in section 5.3 .1 .11 .5 but measure at $S T 4$ ．
－All inputs ST1．b7 to blo＝logic Lo
－Adjust signal generator level so that the pointer of the millivoltmeter is on 2.0 V ．
－Apply logic H to $\mathrm{ST1}$.b 8 and reduce the attenuator setting by 20 dB ． Difference in pointer deflection ．．．．．．．．．．．．．．．．．．．．．．．．．．$\leqq 1 \mathrm{~dB}$

- Apply logic $H$ to ST1.b7 and reduce the attenuator setting by further 20 dB . Difference in pointer deflection $\leqq 1 \mathrm{~dB}$
5.3.1.11.8 Checking the $200-\mathrm{Hz}$ Bandwidth and Ad.justment of Gain
- ST1.all = logic H ( $200-\mathrm{Hz}$ filter ON ).
- Vary the signal generator frequency between the lower and the upper $6-\mathrm{dB}$ cut-off frequency.
$B_{6 ~ d B}$................................................................... $200 \mathrm{~Hz} \begin{aligned} & +20 \mathrm{~Hz} \\ & -30 \mathrm{~Hz}\end{aligned}$
- Tune to the maximum in the pass-band curve.
- Switch over to noise filter:

Note readout on millivoltmeter.

- Switch back to $200-\mathrm{Hz}$ filter:

Adjust for same readout on millivoltmeter by means of R102.

### 5.3.1.11.9 Checking the Overload Detection

Settings on the ESH 3 :
IF bandwidth 2: 500 Hz to 10 kHz (ST1.all = logic L)
IF attenuation 40, 41: 0. dB.
ST1.b11 = logic I if EMF at ST2 20 mV
5.3.1.12 Indication and AF Demodulation Board (Y12)

### 5.3.1.12.1 Supply Voltages

| Pin | Voltage | Current | Frequency |
| :--- | :--- | :--- | :--- |
| ST1.b24 | Ground | - | - |
| ST1.b23 | $+5.25 \mathrm{~V}+0.1 \mathrm{~V}$ | approx. 1 mA | - |
| ST1.a22 | $+10 \mathrm{~V}+0.1 \mathrm{~V}$ | approx. 100 mA | - |
| ST1.b22 | $-10 \mathrm{~V}+0.1 \mathrm{~V}$ | approx. 55 mA | - |
| ST1.ab1 | TIL | - | 500 Hz |

### 5.3.1.12.2 Adjustment of Indication Section

Test setup:
Connect $30-\mathrm{kHz}$ signal generator via attenuator to ST2. As ST2 is a high impedance input, use a $50-\Omega$ termination in parallel.

a) Iinearity of B14:

Settings on the ESH 3: Operating range 33: 20 dB

- DVM check point: pin 6 of B14.
- Input voltage 20 mV ; note reading ( 0.8 .210 mV ).
- Input voltage 2 mV ; vary R164 unt11 $1 / 10$ of the above measured value is obtained ( $e . g$ s $21 \mathrm{mV}+0.5 \mathrm{mV}$ )
= Repeat adjustment to ensure optimum voltage setting.
b) Linearity of B15:

Settings on the ESH 3: Operating range 33: 20 dB

- DVM check point: pin 2 of B15.
- Input voltage 20 mV ; note reading (e.g. 210 mV ).
- Input voltage 2 mV ; vary R174 until $1 / 10$ of the above measured value is obtained (e.g. $21 \mathrm{mV}+0.5 \mathrm{mV}$ )。
- Repeat adjustment to ensure optimum voltage setting.
c) Average-value indication:

Settings on the ESH 3: Operating range 33: 20 dB 35: AV.

- DVM check point STI.22.
- Input voltage 20 mV ; vary Ri82 until the DVM reads ...... $2 \mathrm{~V}+5 \mathrm{mV}$
- Input voltage 2 mV : vary RI77 until the DVM reads ....... $0.2 \mathrm{~V}+5 \mathrm{mV}$
- Repeat adjustment to ensure optimum voltage setting.
d) CISPR indication:

Settings on the ESH 3: 35: CISPR
Frequency between 10 kHz and 149.9 kHz (CISPR 3).

- DVM check point: ST1.a2.
- Input voltage 20 mV ; vary R198 until the DVM reads ...... $2 \mathrm{~V} \pm 5 \mathrm{mV}$
- Input voltage 2 mV ; vary R195 until the DVM reads ........ $0.2 \mathrm{~V} \pm 5 \mathrm{mV}$
- Repeat adjustment to ensure optimum voltage setting.
e) IOG 40:

Settings on the ESH 3: Operating range 33: 40 dB 35: AV

- DVM check point: ST1.a2.
- Input voltage 200 mV ; vary R139 until the DVM reads ..... $2 \mathrm{~V} \pm 5 \mathrm{mV}$
- Input voltage 2 mV ; vary R186 until the DVM reads ....... $0.2 \mathrm{~V}+5 \mathrm{mV}$
- Repeat adjustment to ensure optimum voltage setting.
f) IOG 60:

Settings on the ESH 3: Operating range 32: 60 dB
32: AV

- DVM check point: ST1.a2.
- Input voltage 2 V ; vary R142 until the DVM reads .......... $2 \mathrm{~V}+5 \mathrm{mV}$
- Input voltage 2 mV ; vary R188 until the DVM reads ....... $0.2 \mathrm{~V} \pm 5 \mathrm{mV}$
- Repeat adjustment to ensure optimum voltage setting.
g) Peak-reading time constants:
(With the ESH 3, determined by PEAK and the measuring time)
Settings on the ESH 3: Operating mode 38: GEN. OFF
Oscilloscope check point: B2O pin 10
. $6 V-$

h) 4th oscillator:

Feed TTL-level reference frequency of 500 Hz into $\mathrm{STl} . a b l$. Connect frequency counter to $\mathrm{B} 33 / 2$.

Table 5-9

| Switch position |  | Nominal frequency |  |
| :--- | :--- | :--- | :--- |
| AO | GEN.OFF | 30.0 kHz | $(+10 \mathrm{~Hz})$ |
| A1 | GEN.OFF | 31.0 kHz | $(+10 \mathrm{~Hz})$ |
| USB | GEN.OFF | 31.5 kHz | $(+10 \mathrm{~Hz})$ |
| LSB | GEN.OFF | 28.5 kHz | $(+10 \mathrm{~Hz})$ |
| ANY | GEN.ON | 30.0 kHz | $(+10 \mathrm{~Hz})$ |

Same as above, but connect oscilloscope to $S T 5$ instead of frequency counter.

Settings on the ESH 3: Operating mode 38: GEN.ON
Nominal level $40 \mathrm{mV} \mathrm{rms}=110 \mathrm{mV}_{\mathrm{pp}}=15 \mathrm{dBm}+1 \mathrm{~dB}$ into $50 \Omega$
i) Remote frequency measurement (RFM):

- Connect a signal generator variable between 25 and 35 kHz , approx. 20 mV , to ST2.
- Select REN. FREQ. mode.
- Counter connected to ST5 should read out signal generator frequency.
- Measure nominal level at ST5 by means of oscilloscope: $40 \mathrm{mV} \mathrm{rms}=110 \mathrm{mV}{ }_{\mathrm{pp}}=15 \mathrm{dBm}(+1 \mathrm{~dB})$ into $50 \Omega$ 。


### 5.3.1.12.3 AF Demodulation

a) FM demodulation:

Oscillator adjustment:

- Connect signal generator to ST2: 30 kHz , approx. 20 mV
- DVM check point: ST1.a4
- Vary Ri9 until the DVM reads 0 V ( +20 mV ).

Checking the FM demodulation

- Connect signal generator and DVM as for oscillator adjustment.

Table 5-10

| Signal generator frequency | Nominal voltage reading on DVM |
| :---: | :---: |
| 25 kHz | $-5 \mathrm{~V} \pm 0.3 \mathrm{~V}$ |
| 27 kHz | $-3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ |
| 29 kHz | $-1 \mathrm{~V}+0.3 \mathrm{~V}$ |
| 31 kHz | $1 \mathrm{~V} \pm 0.3 \mathrm{~V}$ |
| 33 kHz | $3 \mathrm{~V} \pm 0.3 \mathrm{~V}$ |
| 35 kHz | $5 \mathrm{~V}+0.3 \mathrm{~V}$ |

b) AM demodulation:

Settings on the ESH 3: AF demodulation 4: A3
Operating mode 38: GEN.OFF
Connect signal generator to $S T 2$ (frequency 30 kHz , level approx. 100 mV , $100 \%$ modulation depth).

Oscilloscope check point ST4:


Vary the input voltage between 2 mV and 2 V . The voltage change at $\mathrm{ST}^{4}$ should be $<3 \mathrm{~dB}$.
c) SSB demodulation:

Settings on the ESH 3: AF demodulation 4: A1 Operating mode 38: GEN.OFF 35: AV

Connect signal generator to $\mathrm{ST2}$ (frequency 30 kHz , level approx. 20 mV , unmodulated).

Oscilloscope check point ST1.a3:
Nominal signal 1 kHz AF , approx. $130 \mathrm{mV} \cdot \mathrm{rms}=360 \mathrm{mV}$ peak-to-peak ( $+50 \%$ ).
d) SSB AGC:

Settings on the ESH 3 as for SSB demodulation, but the level at ST2 is about 100 mV .

Switch off the signal generator. It should take approximately 1 s until the AGC voltage at B8 pin 9 drops.
5.3.1.13 Performance Check of the Indication and AF Demodulation Board Y12

### 5.3.1.13.1 Indication Demodulation

a) Average-value indication:

Feed unmodulated $30-\mathrm{kHz}$ signal $V$ to $S T 2$. The resulting voltages at $S T 1 . a 2$ are as follows:

Table 5-11


Amplitude-modulated signals cannot be directly measured in the $40-\mathrm{dB}$ and $60-\mathrm{dB}$ operating ranges. Use correction curve (Fig. 2-8).
b) Peak-value indication:

If an unmodulated input signal is applied, the same voltage is obtained at ST1.a2 as with averageovalue indication. A $100 \%$ amplitudemodulated signal increases the indicated value by 6 dB as against average-value indication.
c) CISPR indication:

In the frequency range from 10 kHz to 150 kHz , the signal is weighted in accordance with CISPR 3 and from 150 kHz to 30 MHz in accordance with CISPR 1. If unmodulated signals are applied to ST2, the following voltages are obtained at ST1.a2:

Table 5-12

|  | CISPR 1 | CISPR 3 |
| :---: | :---: | :---: |
| $\begin{array}{r} 2 \mathrm{mV} \\ 10 \mathrm{mV} \\ 20 \mathrm{mV} \end{array}$ | $\begin{array}{r} 217 \pm 10 \mathrm{mV} \\ 680 \pm 20 \mathrm{mV} \\ 2170 \pm 10 \mathrm{mV} \end{array}$ | $\begin{array}{r} 200 \pm 10 \mathrm{mV} \\ 630 \pm 20 \mathrm{mV} \\ 2000 \pm 10 \mathrm{mV} \end{array}$ |
| Pulse weighting time constants: |  |  |
|  | CISPR 1 | CISPR 3 |
| Charging time constant Discharging time constant |  | 45 ms <br> 500 ms |

### 5.3.1.13.2 AF Demodulation

a) - Demodulation outputs:

- AF output ST1. 23

Apply 100\% amplitude-modulated, or frequency-modulated (frequency deviation 5 kHz ), input signal.
AF output signal ............................. $200 \mathrm{mV}_{\text {rms }}+50 \mathrm{mV}\left(Z_{\text {out }}=5 \mathrm{k} \Omega\right)$

- AM output ST4

This output is used for measuring the modulation depth by means of an oscilloscope: $100 \%$ modulation depth corresponds to
$1 \mathrm{~V}_{\mathrm{pp}}+3 \mathrm{~dB}\left(\mathrm{Z}_{\text {out }}=1 \mathrm{k} \Omega\right)$.

- FM output ST3

At this output, the frequency offset of the $30-\mathrm{kHz}$ input signal can be measured:
5 kHz deviation corresponds to +0.5 V (tolerance 0.03 V )
$\left(\mathrm{z}_{\text {out }}=10 \mathrm{k} \Omega\right)$.

- Frequency offset output ST1.a4

Here too, the frequency offset of the $30-\mathrm{kHz}$ input signal can be measured: $\pm 5 \mathrm{kHz}$ offset corresponds to $\pm 5 \mathrm{~V}$ (tolerance 0.3 V ) ( $z_{\text {out }}=10 \mathrm{k} \Omega$ )。
b) IF amplifier AGC

AM AGC time constant 0.5 s
AO AGC time constant 0.5 s
A1 hang AGC with fast decay
USB hang AGC
ISB hang AGC
c) 4 th oscillator

| Demodulation mode | Signal <br> generator | Frequency of <br> 4th oscillator |
| :--- | :--- | :--- |
| AM | OFF | switched off |
| FM | OFF | switched off |
| AO | OFF | 30.0 kHz |
| A1 | OFF | 31.0 kHz |
| USB | OFF | 31.5 kHz |
| ISB | OFF | 28.5 kHz |
| any | ON | 30.0 kHz |

d) $30-\mathrm{kHz}$ output $\mathrm{ST5}$

With 30 kHz ON (GEN.ON), the frequency of the output signal is 30.0 kHz . In the REM. FREQ. (remote frequency measurement) mode, the output signal is synchronized with the input signal at ST2.

Level in both cases ...................................... $40 \mathrm{mV}+1 \mathrm{~dB}$ into $50 \Omega$
5.3.1.14 Attenuator Control (Y13)
5.3.1.14.1 Supply Voltages

| Pin | Voltage | Current |
| :--- | :--- | :--- |
| ST3.ab2 | Ground | - |
| ST3.ab1 | $+12 \mathrm{~V}+0.1 \mathrm{~V}$ | approx.2. A |

The Attenuator Control (Y13) is preferably checked together with the RF Attenuator (Y16).

For production testing, it is advisable to use a control arrangement that permits the control levels at ST1.a2 to 16 to be switched on as desired. For checking the gating with B 1 to B 4 at the test points B to K , refer to Table 5-13.
(The attenuation values a are obtained by summing the attenuation values defined by the logic $H$ signals at the control inputs of ST1.)

Table 5-13

| Test point | B | C | D | E | F | H | I | K |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Attenuator/dB | 10 | 20 | 40 | 40 | 10 | 20 | CAL | 1 |
| Attenuation $\mathrm{a} / \mathrm{dB}$ |  |  | . |  |  |  |  |  |
| 0 | L | L | 1 | L | L | L | L | L |
| 10 | H | L | L | 1 | L | $\underline{L}$ | L | L |
| 20 | L | H | L | 1 | L | L | $L$ | 1 |
| 30 | H | H | L | L | L | L | 1 | L |
| 40 | L | L | H | 1 | 1 | $L$ | $\underline{L}$ | 1 |
| 50 | H | L | H | L | L | 1. | L | L |
| 60 | L | H | H | L | L | L | L | L |
| 70 | H | H | H | L | L | L | L | L |
| 80 | L | L | H | H | $L$ | L | L | 1 |
| 90 | H | L | H | H | L | L | L | $\underline{L}$ |
| 100 | L | H | H | H | 1 | L | L | 1 |
| 110 | H | H | H | H | 1 | L | L | $\underline{L}$ |
| 120 | H | H | H | H | H | $L$ | 1 | L |
| 130 | H | H | H | H | L | H | L | 1 |
| 140 | H | H | H | H | H | H | L | L |
| CAL | H | L | H | H | x | x | $\mathrm{H}_{*}$ ) | X |
| 1 | X | X | X | X | X | X | ${ }^{\text {x }}$. | H |

*) No effect, i.e. the last setting is preserved.

It is possible to set attenuation values up to 140 dB in the ESH 3. If, when the ESH 3 is being repaired, there is any doubt as to the proper functioning of the Attenuator Control, it is recommended that the RF Attenuator be removed from the ESH 3, the base plate unscrewed and the functioning of the control be checked by observing the switching state of the contacts (for location of the attenuators, see circuit diagram 303.2813 S ). An error can be tracked down with the aid of section 4.1.11.

### 5.3.1.15.1 Access to Circuirs

For troubleshooting, the instrument may be operated with the power supply swung out. For this the instrument panelling is first removed. After then removing six screws on the frame, the power supply can be withdrawn horizontally. After the actuating rod of the power switch is pulled off, the power supply can be placed towaxd the back on the cooling body.

Note! In dismounting the power supply, only the six outermost screws may be removed, since otherwise other parts of the power supply will be loosened.

In remounting the power supply in the instrument, the flat cable must be correctly folded and the actuating rod of the power swicch replaced before the power supply is moved into place and screwed to the frame.

The individual subassemblies of the power supply are best tested while connected into the instrument cixcuit, since all supply and control lines are then attached. The subassemblizes can be swivelled apart to pemit access to the components. So long as the power supply is operated for only short periods or is only partially loaded, the large heat sinks can be removed. After the heat sinks of the analog power supply have been unscrewed, all components and test points are accessible. The solder side of the subassembly is accessible after the latter has been unscrewed. The removed subassembly must be supported in a suitable fixture to assure that no shorecircuits are possible.

For measurements on the solder side of the swicching power supply board, the cover on the side of the analog power supply must be unscrewed.
The analog board remains attached to the cover. For replacement of components in the switching power supply, the boardis not removed, instead the entire frame is unscrewed from the rear panel. For this the two large screws on the rear side in the cooling fins and then the six small screws in the frame be removed. The frame can then be swivelled away without having to remove the leads to the rear panel. The rectifier plate must remain in place for operation of the system.

Before reassembly, the correct seating of the socket strip on $X 30$ and the lines on the lead-through filters must be checked. The heat sink on the switching power supply must be provided with heat-conducting paste. Then the frame screws and screws in the cooling fins are loosely screwed in. After lining up the unit the frame is first tightened and then the internal heat sink.

Note! If the internal heat sink is not tightly screwed down, the switching power supply will overheat in operation.

### 5.3.1.15.2 Adjustment of Reference Voltages

The only adjustment points of the power supply are potentiometers R89 and R95 on the analog power supply board, with which the monitoring and regulator reference voltages can be set.

Note! : This adjustment should only be undertaken if really necessary, since other circuits will then also have to be adjusted.

Reference point for all accurate voltage measurements is the ground neutral point XS on the rectifier board. The adjustment is made by measuring the $+10-\mathrm{V}$ supply voltage at a point of the sensing line (motherboard X201A4) and accurately setting it with R95. If the accessories are available, the $+10-V$ supply can also be adjusted by measuring this voltage on the sensing line at points X 18.16 or X 36.1 of the analog power supply board. After this adjustment, the regulator reference voltage, measurable on test connector X9.1, must have the value $+8 \mathrm{~V} \pm 20 \mathrm{mV}$. The monitoring reference voltage at X 9.3 must be set to exactly the same value by means of R89.

The other output voltages are determined by fixed resistors and cannot be separately adjusted. The $-10-V$ supply at sensing point X201B4 on the motherboard tracks the $+10-\mathrm{V}$ supply with a maximum error of 10 mV . All other voltages must lie within a tolerance limit of $\pm 2 \%$ at the power supply. It must be noted that because of voltage drops in the lines, the output of the $+5-\mathrm{V}$ supply is 5.5 V at its output terminals. The output voltage of the $+30-\mathrm{V}$ supply lies between 31 V and 40 V depending on the loading of the $-11-\mathrm{V}$ supply. When the $-11-\mathrm{V}$ supply is not loaded, the output of the $+30-\mathrm{V}$ supply can drop so low that the H3 LED lights. This is of no significance in normal operation.

The analog power supply has a variety of connectors, some of which are bridged with shorting links.

```
X31 to X38 . 1-.2 connected (comparator inputs)
X5 . 2-.3 connected (short-circuit link right)
```

X9, X10
X2,X4,X8, X12

```
open
\(.1-.2\) connected for setting 20 V
(short-circuit top, ESVP)
.2-. 3 connected for setting 25 V
(short-circuit bottom, ESH3)
```


### 5.3.1.15.4 Check of Output Currents

In table 5-2 are listed the maximum output currents of the power supply these may, however, not be drawn in operation. The measured current values must lie within the colerance range of $-10 \%$ to $+25 \%$.

Table 5-14a Output Currents

| Voltage <br> (nominal) | Maximum current <br> cold | wharm | Shortcircuit current |
| :--- | :--- | :--- | :--- |
| +5 V | 7 A | 6 A |  |
| +12 V | 4 A | 3 A | 6 A |
| -11 V | 1.4 A | 1.2 A | 3 A |
| +33 V | 0.5 A | Can only be overloaded for |  |
|  |  | 1.2 A |  |
| +10 V | 0.7 A |  |  |
| -10 V | 1.3 A |  | 0.15 A |
| +20 V (ESVP) | 0.4 A | 0.33 A |  |
| +25 V (ESH3) | 0.05 A | 0.08 A |  |
| +30 V | 0.06 A |  | 0.01 A |

The check on analog supplies with foldback current limiting can only be made with resistors of corresponding power rating. The switching supplies can also be checked with electronic current sinks.

The sensing leads for the $+10-V$ and $-10-V$ supplies must be connected to the proper outputs. If these leads are not connected, the supply voltages will be about 0.3 V too high, although the voltages on the comparators are correct.

### 5.3.1.15.5 Cheek of Monitoring Circuit

When the shorting links are removed from X31 to X38, the corresponding LEDs H1 to H8 light. By applying a variable voltage to the pins. 2 , the limit values of the monitoring windows can be determined. The actual supply voltages can be measured at the same time on pins .1.

Note! The test with X 31 may only be made if the internal supply is operating properly, since otherwise the other supplies can be switched into prohibited states.

Because of the interlocking of some of the power supplies, only one external voltage at a time must be applied and the corresponding LED observed. In case of automatic measurement the error messages are monitored at the outputs of these messages and on S1, S2, S3.

Table 5-14b Monitoring Window Data

| Connector | LED | Nominal voltage | Lower limit | Upper <br> 1imit | Max. shift of window |
| :---: | :---: | :---: | :---: | :---: | :---: |
| X31 | H1 | 15 V | 12.8 V | $18 \mathrm{~V}$ <br> No mess. | 300 mV |
| X32 | H2 | 24 V | $\begin{aligned} & \$ 20.8 \mathrm{~V} \\ & \hat{\uparrow} 21.3 \mathrm{~V} \end{aligned}$ | $40 \mathrm{~V}$ <br> No mess. | 400 mV |
| X33 | H3 | 30 V | 28.8 V | 31.4 V | 600 mV |
| X34 | H4 | 25 V (ESH3) | 23.9 V | 26.1 v | 500 mV |
| $\times 34$ | H4 | 20V (ESVP) | 19.1 V | 20.9 V | 400 mV |
| $\times 35$ | H5 | 12 V | 11.4 V | 12.5 V | 200 mV |
| X36 | H6 | 10 V | 9.9 V | 10.1 V | 20 mV |
| X37 | H7 | -10 V | -9.9 V | -10.1 V | 20 mV |
| $\times 38$ | H8 | 5 V | 5.1 V | 5.6 V | 50 mV |

Table 5-14c Message Output Levels

| Connector | Nominal voltage | LED | Output on Pin | $\begin{aligned} & \text { Leve } \\ & 0 . \mathrm{K}_{1} \end{aligned}$ | el Error | Other responses in case of error |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\times 31$ | 15 V | H1 | S1 818.23 | 0 V | $>0.8 \mathrm{~V}$ | All the others |
| X 32 | 24 V | H2 | S3 X18.25 | 0 V | 20.8 V | All except $51, \$ 2,12 \mathrm{~K}$ |
| X33 | 30 V | H3 | 30 comp. x14. 15 | high | low |  |
| $\times 34$ | $20 / 25 \mathrm{~V}$ | H4 | 20 comp. X14.18 | high | low |  |
| $\times 35$ | 12 V | H5 | 12 comp. X14.17 | high | 10w |  |
| X36 | 10 V | H6 | 10 comp. X14.9 | high | low |  |
| $\times 37$ | $-10 \mathrm{~V}$ | H7 | -10 comp. X14.16 | high | low |  |
| X38 | 5 V | H8 | S2 X18.26 | 0 V | $>0.8 \mathrm{~V}$ | All except S 1 |

The ERROR INT, TRAP and RESET outputs can be monitored simultaneously. If all voltages are in order, ERROR INT and TRAP are Low and RESET is high. ERROR INT goes high at every error.

TRAP is high during the on-time of D4I (about 100 ms ) if S 3 responds.
RESET goes low when $S 2$ responds or after TRAP has gone back to low.
RESET then goes high again only after monostable D $4 \bar{I} \bar{I}$, having been triggered by the negative edge of ERR INT, returns to its off-state.

The output of the $5-V$ comparator and of monostable D4II can be monitored on X10.

### 5.3.1.15.6 Notes on Troubleshooting

If the green LED in the power supply does not light, the microprocessor is blocked (by an ERROR message or RESET). The fault can be localized with help of the description for the monitoring circuit in Section 5.1.4. In any case, before any repair is attempted, it should be determined whether the error is the result of external influence (undervoltage, overload).

No LED is lit
If after instrument switch-on no LED in the power supply lights, the cause quite likely is that there is no input to the analog power supply. After check of the $A C$ supply or battery voltage and the fuses accessible on the rear panel, the voltage on capacitor $C 5$ of the rectifier board should be checked with the $A C$ power supply on. Flat connectors $X 1$ and $X 2$ are accessible from the underside of the instrument after the paneliing is removed. If a voltage is present, the power supply should be removed from the instrument and the connection to the analog power supply checked. The input voltage should also be present on regulator N5. 1 of the analog power supply.

One or more red LEDs 1it
The rank order of the error messages as shown in Fig. 5-2 must be observed, i.e. the higher ranking signal must be checked first. If the green LED lights when the load on the power supply is reduced or completely removed, the output currents should be checked against the values in Section 5.2.4.

Before attempting any repairs, the voltages in question and the reference voltages should be checked to eliminate the possibility of a fault in the monitoring circuit. Before opening the switching power supply, the turn-off signals S1, S2, S3 (accessible on comnector X17 of the rectifier board) should be checked. If a volrage $>0.8 \mathrm{~V}$ is present, the corresponding switching supply is turned off.

After unscrewing the cover of the power supply, the voltages on connector X30 should be checked, in order to eliminate the possibility of a break in the connection between the switching power supply and the rectifier board. In this condition the signals on all components of the switching power supply can be checked.


Fig. 5-6 Rank order of supply voltages and error messages

### 5.3.1.16 Checking the Recorder Control Board

(See circuit diagram 335.9913 S)

### 5.3.1.16.1 IEC-bus Section

- The IEC bus is simply connected via a flat cable (ST8) from the Computer board (BU8) to the recorder control board. For checking, unplug SI8 and use a continuity tester.

NOLE: If the recorder control board has to be removed, protect the pins of plugs 7 and 8 against danage by sticking them in pleces of foam plastic.

- Pull out ST7 from BUT.
- Connect logic tester probe (or voltmeter) to $\$ 27$ pin .
- Switch on receiver and apply +5 V. Check function of the address line (al).
- Change connections and check a2 to 5 and ton siailariy.
- Conrect logie tester probe or voltmeter to ST7 pin 7.
- If BUI pir 1 is shortcircuited to ground. ST7 pin 7 should be at logic $I$ ( $<0.8 \mathrm{~V}$ ).
- Check ST7 pin 8 with BUR pin 2 to ground, as above.
- Check the remaining pins of SI7, by means of a continuity tester sgainst the circuit diagram.


### 5.3.1.16.2 Checkins the Demultiplexer

- Unplug STT.
- Switch on the ESH 3 .
- Apply a stable voltage $v=0$ to 10 V (E.g. 5 V) to 66 pin 5.
- Deliver logic I to ST7 pin 13 (penist signai).
- Checi in accordance with Table 5-15:

Table 5-15

| Fecorder | $\begin{array}{lll} \text { SI7 pin } & \\ 12 & 11 & 10 \end{array}$ | $\begin{aligned} & \text { Bue pin } \\ & 8 \quad 9 \end{aligned}$ |  |  |  |  |  |  |  | 24 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | H 2 L | $v_{1 n} x$ | $x$ | x |  | $\pm$ | \% | H | H | 8 |
| 2 | L H L | $x \quad v_{1 n}$ | $x$ | $x$ |  | H | $L$ | H | H | H |
|  | H H L | $x \quad x$ | $v_{1 n}$ | $x$ | $x$ | H | H | L | н | H |
|  | L L H | $x \quad x$ | $x$ | $v_{1 n}$ | $x$ | H | H | H | $\pm$ | H |
| 5 | $\begin{array}{lll}\text { H } & \text { L }\end{array}$ | $x \quad x$ | $x$ | $x$ | $\nabla_{18}$ |  |  |  |  | 1 |
| 0.6 \% 8 | Recordars not aetive |  |  |  |  | +) | 2 l : | Po | 118 t |  |

### 5.3.1.16.3 Checking the External Reference Switchover

- Check the following levels using a voltmeter or logic tester probe:

| Synthesizer 2/pin | Inc. ref. | Ext. 5 MHz | Ext. 10 MHz |
| :--- | :--- | :--- | :--- |
| ST1/a2 | L | H | L |
| ST1/b2 | L | L | H |

- Check the pulling range of the synthesizer in accordance with 5.3.1.3.4.3.


### 5.3.1.17 Motherboard (Y18)

Check the Motherboard in accordance with circuit diagram 303.2020 S.

### 5.3.1.18 RF Attenuator (Y16)

It is recommended that the RF attenuator Y16 be checked together with the attenuator control Yi3.

Note: Torque range for SMA socket: 80 to 120 Ncm

- Screw down the six adjustment screws of the base plate all the way.
a) Check the residual attenuation with $D C$ voltage.
- Set the level switch to 0 dB .
- Measure the resistance between the inner conductor of the input and the inner conductor of the output by means of an ohmmeter with a resolution of $100 \mathrm{~m} \Omega$ 。
- The resistance consists of the contact resistances of the 18 switching contacts and the series resistances of the thinmilm conductors. The resistance of the test setup must be taken into account. The resistance $\mathrm{R}_{\mathrm{T}}$ should be $\leq 800 \mathrm{~m} \mathrm{~m}_{0}$ The resulting residual attenuation $\mathrm{a}_{\mathrm{O}}$ is given by the formula

$$
a_{0}=20 \log \frac{100}{100+R_{T} / \Omega} \text { and is }<0.07 \mathrm{~dB} \text {. }
$$

- Then measure the resistance between the inner conductor of the calibration input and the inner conductor of the output. To do so, apply a short switching pulse across the coil of solenoid 7 which initiates calibration of the RF attenuator without switching the $10-\mathrm{dB}$ attenuator pad into circuit. The resistance should be $<500 \mathrm{~m} \Omega$
b) Checking the attenuator pads with a $D C$ voltage

Test setup:


- Connect a constant DC voltage source $V \leqq 5 V$ with an internal resistance of $50 \Omega$ to the input of the RF attenuator.
- Measure the output voltage of the through-connected ( 0 dB ) attenuator by means of a digital voltmeter. The attenuator is terminated with $50 \Omega$.
- Successively set the attenuation values $a=1,4,10,20$ and 40 dB . The attenuator control cannot be used for this purpose; instead, short $12-V$ switching pulses must be applied directly to the terminals accessible on the control board. Open the connection to the attenuator control.

The attenuation actually provided can be calculated from the voltage ratios:

$$
a=20 \log \frac{V(0 d B)}{V(a d B)}
$$

The maximum permissible attenuation error can be read off the below table:

| Attenuator pad/dB | $\mathrm{a}_{\mathrm{min} / \mathrm{dB}}$ | Actual/dB | $\mathrm{a}_{\mathrm{max} / \mathrm{dB}}$ |
| :---: | :---: | :---: | :---: |
| 1 | 0.98 |  | 1.02 |
| 4 | 3.98 |  | 4.02 |
| $10(1)$ | 9.96 |  | 10.04 |
| $10(2)$ | 9.96 |  | 10.04 |
| $20(1)$ | 19.94 |  | 20.06 |
| $20(2)$ | 19.94 |  | 20.06 |
| $40(1)$ | 39.92 |  | 40.08 |
| $40(2)$ | 39.92 |  | 40.08 |

The location of the individual attenuator pads is shown in circuit diagram 303.2813 S .

Since the RF attenuator is used only up to 30 MHz in the ESH2 even though it is designed for frequencies up to 2.7 GHz , there is no need for an RF check (VSWR < 1.2 at input and output in the range 0 to 1000 MHz ).

### 5.3.2 Overall Adjustment

Prerequisites for the adjustment:

- All assemblies must be operational and adjusted as required.
- Remove Filter 2 (Y8), open it and connect it to the receiver via the Service Adapter.


### 5.3.2.1 Checking the Supply Voltages

Motherboard b23 ............................................... $4 .+250 \mathrm{mV}$
Motherboard a22 .............................................. $10.000 \mathrm{~V}+200 \mathrm{mV}$
Motherboard a23 ............................................ $+12 \mathrm{~V} \pm 500 \mathrm{mV}$
Motherboard 224 ............................................... $25 \mathrm{~V} \pm 130 \mathrm{mV}$
Filter Control Y6, a24 .................................... $30 \mathrm{~V} \pm 1.35 \mathrm{~V}$
Motherboard b22.............................................. $10 \mathrm{~V} \pm 400 \mathrm{mV}$

### 5.3.2.2 Checking the Gain of Mixers $1+2$ (Y9) and Mixer 3 (Y11)

Settings on the ESH 3:

| Indicating mode | 35: AV. |
| :--- | :---: |
| IF bandwidth | $6: 500 \mathrm{~Hz}$ |
| IF artenuation | $40,41: 40 \mathrm{~dB}$ |
| RF atrenuation | $40,41: 50 \mathrm{~dB}$ |

Connect signal generator ( $f<10 \mathrm{MHz}$, level $80 \mathrm{~dB} \mu \geqslant \neq 0.1 \mathrm{~dB}, 50 \Omega$ ) to the RF input 45 of the ESH 3. Adjust the gain by means of R20 on Yll so that at a medium gain setting ( $+2 \mathrm{~V}+0.2 \mathrm{~V}$ ) at the base of Tl (Y11) full-scale deflection is obtained.

### 5.3.2.3 Checking for Equal Gain with the Various Bandwidths

Settings on the ESH 3: See section 5.3.2.4
Permissible difference in the indicated voltage when selecting a new bandwidth ................................................ ${ }^{1} \mathrm{~dB}$

Reference: $500-\mathrm{Hz}$ bandwidth
Adjust the gain with
$200-\mathrm{Hz}$ bandwidth: on Y11 using R102
$2.4-\mathrm{kHz}$ bandwidth: on Y9 using R69
$10-\mathrm{kHz}$ bandwidth: on Y9 using R76.

### 5.3.2.4 Adjusting Filters 15 and 16 (Y8)

Connect the input of the receiver to a VSWR meter which can be driven from the GEN. output of the ESH 3.
a) Adjusting filter 15 for minimum reflection coefficient

Table 5-16

| $f$ | by means of |
| :--- | :---: |
| R.... on $Y 6$ |  |
| 10.0 MHz | 79 |
| 11.0 MHz | 55 |
| 15.0 MHz | 70 |
| 19.0 MHz | 89 |
| 19.9 MHz | 103 |

Repeat this adjustment at least once, since the settings interact.
b) Adjusting filter 16 for minimum reflection coefficient

Adjust L78, L79, C162 and C164 in the frequency range from 20.0 MHz to 29.9 MHz .

After the adjustment fasten the coils and secure the windings with a suitable varnish or glue. The board can now have the cover fitted and be inserted in the unit.

Checking the input impedance:
Filters 1 to 14: at least three readings per filter range
Filters 15 and 16: at least every 2 MHz .
VSWR with $0 \mathrm{~dB} R F$ attenuation ................................ $2=<33 \%$
with RF attenuation $>0 \mathrm{~dB}$............................. $1.2=<10 \%$

### 5.3.2.5 Checking the Reference Frequency

Tune the receiver to 29.0000 MHz . Switch on the generator and connect to a frequency counter with a frequency accuracy of better than $1 \times 10^{-8}$. After a warm-up period of $>5 \mathrm{~min}$, adjust the counter display to 29.0000 MHz by means of R22 on Y1.
5.3.2.6 Adjustment of Calibration
a) Sinewave calibration

Settings on the ESH 3: Indicating mode 35: AV.

| Operating range | 33: | 20 dB |
| :--- | ---: | :--- |
| IF bandwidth | $\underline{6}:$ | 10 kHz |
| RF attenuation | 40, | 41: |
| IF attenuation | 40 dB |  |
| 41: | 40 dB |  |
| Frequency | $\underline{20}:$ | 1 MHz |
| Operating mode | 38: | TWOPORT |
| Meas time | 37: | $\geqq 0.1 \mathrm{~s}$ |

Connect suitable precision level meter ( $50 \Omega$ ), such as specially calibrated URV 4 with $50-\Omega$ insertion unit, to the GEN. output 44.

- Set level by means of R84 on Y10. Required level ..................................... $80 \mathrm{dBu} \mathrm{V}+0.1 \mathrm{aB}$ $=-27 \mathrm{dBm}+0.1 \mathrm{~dB}$
- Press CAL. button 15 momentarily (calibration check).
- Connect the GEN. output 44 to the RF input 45. Required indication ................................ $0 \mathrm{~dB} \quad+0.1 \mathrm{~dB}$
b) Pulse calibration (CISPR 3)
over frequency range 10 kHz to 149.9 kHz
Settings on the ESH 3: Indicating mode 35: CISPR

| $R F$ attenuation | 40, | 41: |
| :--- | ---: | :--- |
| IF attenuation | 30 dB |  |
| Frequency | $\underline{40}, \underline{41}:$ | 40 dB |
|  | $\underline{20}:$ | 0.1 MHz |

- Press CaL. button 15 and hold down (total calibration).
- Feed sinewave signal ( $0.1 \mathrm{MHz}, 60 \mathrm{~dB} \mu \mathrm{~V}+0.1 \mathrm{~dB}, 50 \Omega$ ) into the ESH 3. Nominal indication after fine tuning to maximum indication $60 \mathrm{~dB} \mathrm{\mu V} \pm 1 \mathrm{~dB}$

Correct, if necessary, by means of R89 on Y10:
Indicated value too high $\rightarrow$ turm R89 counterclockwise $\rightarrow$ calibrate
$\rightarrow$ measure
Indicated value too low $\rightarrow$ turn $R 89$ clockwise $\rightarrow$ calibrate $\rightarrow$ measure.

### 5.3.2.7 Checking the CISPR 3 Weighting Circuit <br> Test item:



Connect a pulse generator with an adjustable repetition frequency in compliance with CISPR 3 requirements to the input of the ESH 3. Feed a pulse with a repetition frequency of 25 Hz , a period of 250 ns and an amplitude of 2.7 V corresponding to the CISPR 3 standard pulse

$$
V \cdot t=\text { approx. } 1.35 \mu V s \text { (ENE) }
$$

into the input of the ESH 3.
Required indication $40 \mathrm{~dB} \mathrm{\mu V}+1 \mathrm{~dB}$

When varying the repetition frequency in compliance with the CISPR 3 requirements, the indication on the receiver must be within the CISPR 3 tolerance limits. (Fig. 5-5).
5.3.2.8 Pulse Calibration (CISPR 1) over Frequency Range $>150 \mathrm{kHz}$

| Settings on the ESH 3: | Indicating mode | 35 | $:$ | CISPR |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  | RF attenuation | $\underline{40}, \underline{41}:$ | 30 dB |  |
|  | IF attenuation | $\underline{40}, \underline{41}:$ | 40 dB |  |
|  | Frequency | $\underline{20}$ | $:$ | 1 MHz |

- Press CAL. button 15 and hold down (total calibration).
- Feed sinewave signal ( $1 \mathrm{MHz}, 60 \mathrm{dBuV}+0.1 \mathrm{~dB}, 50 \Omega$ ) into the ESH 3.

Nominal indication ....................................... $60 \mathrm{~dB} \mu \mathrm{~V}+1 \mathrm{~dB}$
Correct, if necessary, by means of R91 on Y10:
Indicated value too high $\longrightarrow$ turn R91 counterclockwise $\longrightarrow$ calibrate $\longrightarrow$ measure
Indicated value too low $\longrightarrow$ turn R91 clockwise $\longrightarrow$ calibrate $\longrightarrow$ measure.

### 5.3.2.9 Checking the CISPR 1 Weighting Circuit

Same test setup as for CISPR 3.
Standard pulse: V $\cdot t=0.316 \mu \mathrm{Vs}$ (EMF).
This standard pulse with a repetition frequency of 100 Hz should
give an indication of ....................................................... $60 \mathrm{~dB} \mu \mathrm{~V} \pm 1 \mathrm{~dB}$
The calibration pulse generator from Schwarzbeck, Type IGU 2912 supplies a CISPR 2, 4 pulse

$$
V \cdot t=0.044 \mu V \mathrm{~S} \text { (EMF). }
$$

This corresponds to a difference of $-17.12 d B$ from the CISPR 1 pulse. The required indication on the ESH 3 must therefore be $\ldots \ldots . .43 \mathrm{~dB} \mu \mathrm{~V}+1 \mathrm{~dB}$ with the output attenuator of the IGU 2912 set to 60 dB .

When varying the repetition frequency in compliance with the CISPR I requirements, the indication on the receiver must be within the CISPR 1 tolerance limits (Fig. 5-6).

### 5.3.3 Final Check of the Receiver Performance Data

Check the logic functions in accordance with section 2.2 .4 (operating manual) and the performance specifications in accordance with section 3.2 (service manual).

The easy-to-service ESH 3 is designed to minimize the time required to repair it. However, defective components must be replaced only with the parts designated in the relevant parts lists. Certain restrictions must be observed with the following components:

- The detector diodes GI9 and GL1O on Y10 affect the measurement accuracy and the temperature dependence of the receiver. When a repair becomes necessary, the temperature compensation must be readjusted.
- The attenuator Y16 is a thin-film unit whose components can only be replaced by the manufacturers. No attempt should, therefore, ever be made to repair it. The complete defective attenuator must be replaced.

If a great number of ESH 3 s are used at one and the same location, it is recommended that a set of receiver modules be kept as replacement to minimize the down-time of the receiver concerned. The replacement of a receiver module is uncritical as the interfaces within the receiver are precisely defined. (Exception: Filter Control and Filter 2 must be replaced together.)

### 5.5 Mechanical Repair

Since the receiver contains practically no mechanical parts that are subject to mechanical wear and tear - with the exception of the tuning knob and the carrying handle - little mechanical repair will be required.

A normal service tool kit will do for any mechanical repair work that may become necessary.

### 5.6 Spare Parts

The parts lists of the various receiver modules in the appendix contain the manufacturer's identification numbers ( $=$ order numbers) of all the electrical components used. When replacing ICs, make sure to use an IC from the same manufacturer, if possible, since the differences are considerable in spite of the same designation.

To achieve optimum reliability, the boards and modules used in the receiver undergo rigid quality control prior to final assembly.
Components from other makers, such as resistors, capacitors, diodes, transistors, ICs and displays are required to comply with Res specifications ensuring optimum reliability. It is therefore recommended that defective components should only be replaced by original components that have been passed by Res.

